

## 3-3/4 Digit A/D Converter with Frequency Counter and Logic Probe

### Features

- Multiple Analog Measurement System
  - Digit A/D Converter
  - Frequency Counter
  - Logic Probe
- Low Noise A/D Converter:
  - Differential Inputs: (1pA Bias Current)
  - On-Chip 50ppm/°C Voltage Reference
- Frequency Counter:
  - 4MHz Maximum Input Frequency
  - Auto-Ranging Over Four Decade Range
- Logic Probe:
  - Two LCD Annunciators
  - Buzzer Driver
- 3-3/4 Digit Display with Over Range Indicator
- LCD Display Driver with Built-in Contrast Control
- Data Hold Input for Comparison Measurements
- Low Battery Detect with LCD Annunciator
- Under Range and Over Range Outputs
- On-Chip Buzzer Driver with Control Input
- 40-Pin Plastic DIP, 44-Pin Plastic Flat Pack, or 44-Pin PLCC Packages

### General Description

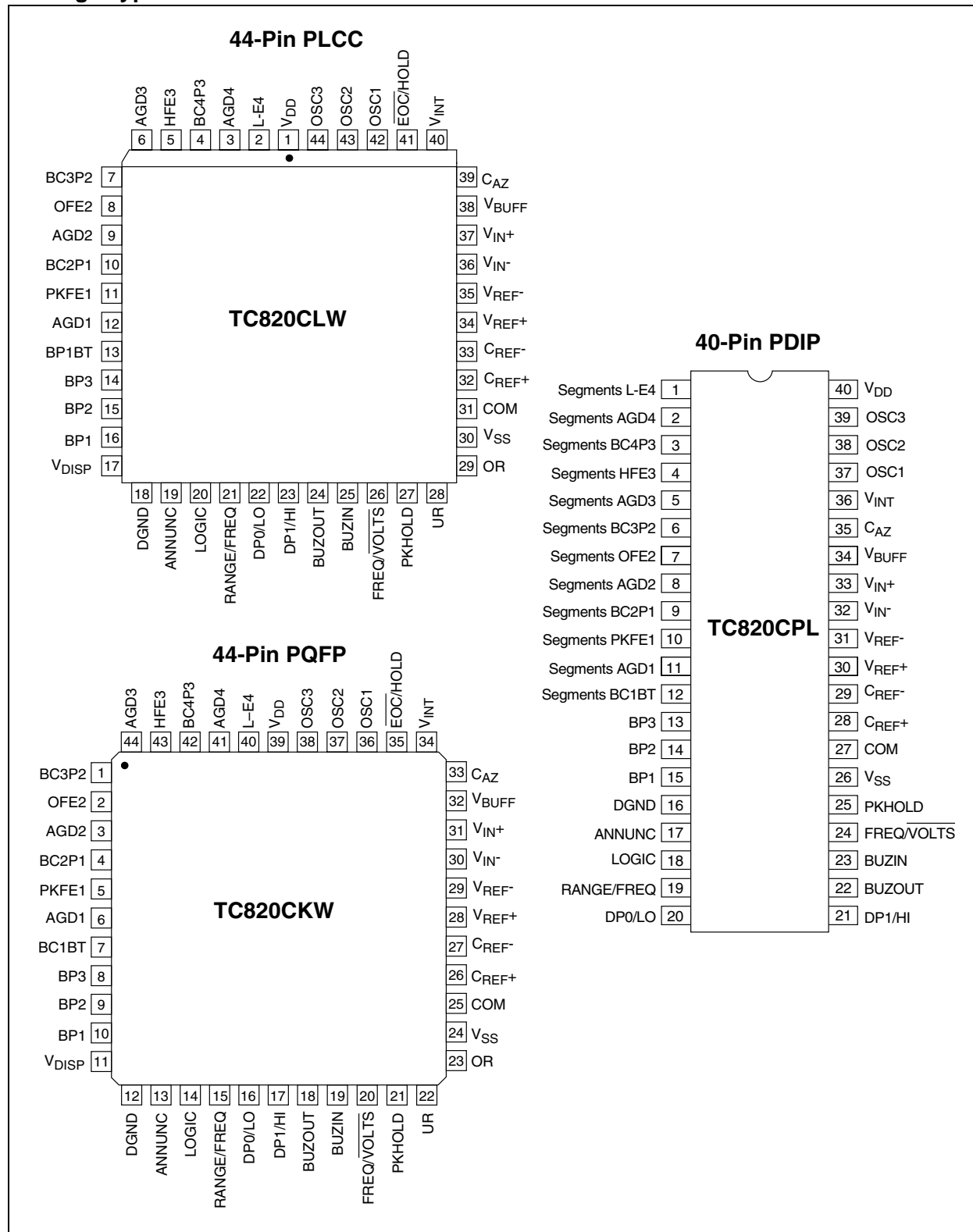
The TC820 is a 3-3/4 digit, multi-measurement system especially suited for use in portable instruments. It integrates a dual slope A/D converter, auto-ranging frequency counter and logic probe into a single 44-pin surface mount, or 40-pin through hole package. The TC820 operates from a single 9V input voltage (battery) and features a built-in battery low flag. Function and decimal point selection are accomplished with simple logic inputs designed for direct connection to an external microcontroller or rotary switch.

### Device Selection Table

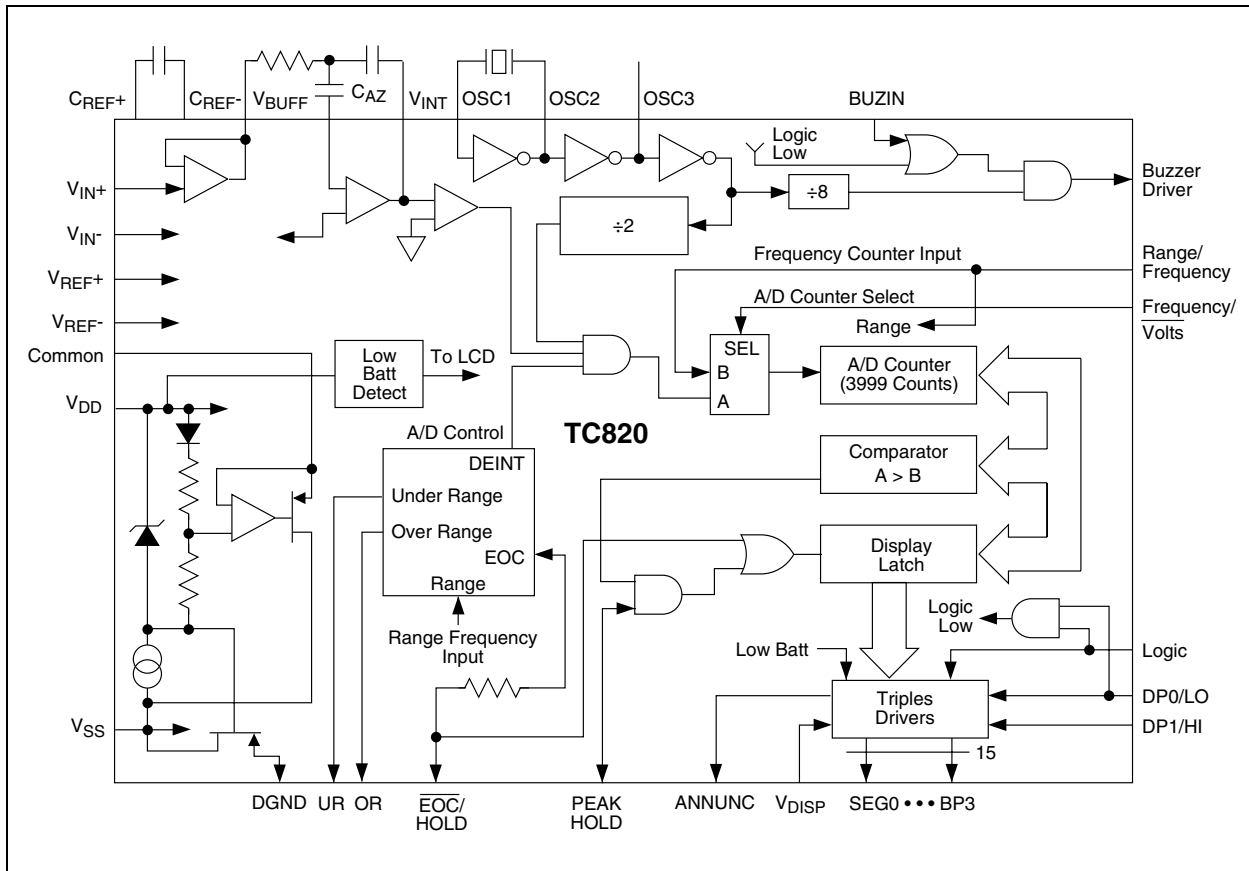
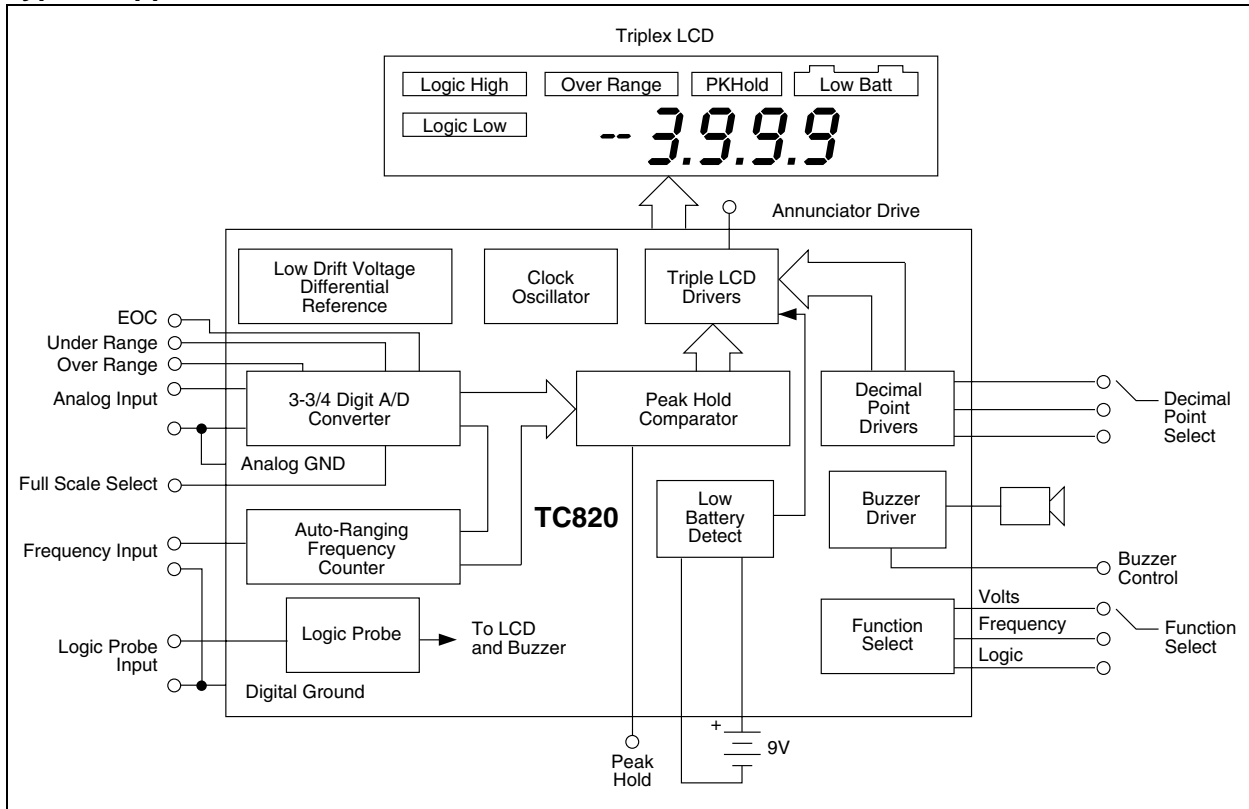
Part Number	Resolution	Package	Operating Temp. Range
TC820CPL	3-3/4 Digits	40-Pin PDIP	0°C to +70°C
TC820CKW	3-3/4 Digits	44-Pin PQFP	0°C to +70°C
TC820CLW	3-3/4 Digits	44-Pin PLCC	0°C to +70°C

# TC820

## Package Type



## Typical Applications



# TC820

## 1.0 ELECTRICAL CHARACTERISTICS

### Absolute Maximum Ratings\*

Supply Voltage ( $V_{DD}$  to GND) ..... 15V

Analog Input Voltage:

(Either Input) (**Note 1**) .....  $V_{DD}$  to  $V_{SS}$

Reference Input Voltage (Either Input) .....  $V_{DD}$  to  $V_{SS}$

Digital Inputs .....  $V_{DD}$  to DGND

$V_{DISP}$  .....  $V_{DD}$  to (DGND – 0.3V)

Package Power Dissipation ( $T_A = 70^\circ\text{C}$ ) (**Note 2**):

40-Pin Plastic DIP ..... 1.23W

44-Pin PLCC ..... 1.23W

44-Pin Plastic Flat Package (PQFP) ..... 1.00W

Operating Temperature Range:

"C" Devices .....  $0^\circ\text{C}$  to  $+70^\circ\text{C}$

"E" Devices .....  $-40^\circ\text{C}$  to  $+85^\circ\text{C}$

Storage Temperature Range .....  $-65^\circ\text{C}$  to  $+150^\circ\text{C}$

\*Stresses above those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only and functional operation of the device at these or any other conditions above those indicated in the operation sections of the specifications is not implied. Exposure to Absolute Maximum Rating conditions for extended periods may affect device reliability.

## TC820 ELECTRICAL SPECIFICATIONS

Electrical Characteristics: $V_S = 9\text{V}$ , $T_A = 25^\circ\text{C}$ , unless otherwise specified.						
Symbol	Parameter	Min	Typ	Max	Units	Test Conditions
	Zero Input Reading	-000	$\pm 000$	+000	Digital Reading	$V_{IN} = 0\text{V}$ Full Scale = 400mV
RE	Rollover Error	-1	$\pm 0.2$	+1	Counts	$V_{IN} = \pm 390\text{mV}$ Full Scale = 400mV
NL	Nonlinearity (Maximum Deviation From Best Straight Line Fit)	-1	$\pm 0.2$	+1	Count	Full Scale = 400mV
	Ratiometric Reading	1999	1999/2000	2000	—	$V_{IN} = V_{REF}$ , TC820
CMRR	Common Mode Rejection Ratio	—	50	—	$\mu\text{V/V}$	$V_{CM} = \pm 1\text{V}$ , $V_{IN} = 0\text{V}$ Full Scale = 400mV ( $V_{FS} = 200\text{mV}$ )
VCMR	Common Mode Voltage Range	$V_{SS} + 1.5$	—	$V_{DD} - 1$	—	Input High, Input Low
$e_N$	Noise (P-P Value Not Exceeded 95% of Time)	—	15	—	$\mu\text{V}$	$V_{IN} = 0\text{V}$ Full Scale = 400mV
$I_{IN}$	Input Leakage Current	—	—	—	—	$V_{IN} = 0\text{V}$
		—	1	10	pA	$T_A = 25^\circ\text{C}$
		—	20	—	pA	$0^\circ\text{C} \leq T_A \leq +70^\circ\text{C}$
		—	100	—	pA	$-40^\circ\text{C} \leq T_A \leq +85^\circ\text{C}$
$V_{COM}$	Analog Common Voltage	3.15	3.3	3.45	V	$25\text{k}\Omega$ between Common and $V_{DD}$ ( $V_{SS} - V_{COM}$ )
$V_{CTC}$	Common Voltage Temperature Coefficient	—	—	—	—	$25\text{k}\Omega$ between Common and $V_{DD}$
		—	35	50	ppm/ $^\circ\text{C}$	$0^\circ\text{C} \leq T_A \leq +70^\circ\text{C}$
		—	50	—	—	$-40^\circ\text{C} \leq T_A \leq +85^\circ\text{C}$

**Note 1:** Input voltages may exceed the supply voltages provided that input current is limited to  $\pm 100\mu\text{A}$ . Current above this value may result in invalid display readings, but will not destroy the device if limited to  $\pm 1\text{mA}$ .

**Note 2:** Dissipation ratings assume device is mounted with all leads soldered to printed circuit board.

## TC820 ELECTRICAL SPECIFICATIONS (CONTINUED)

Electrical Characteristics: $V_S = 9V$ , $T_A = 25^\circ C$ , unless otherwise specified.						
Symbol	Parameter	Min	Typ	Max	Units	Test Conditions
TC <sub>ZS</sub>	Zero Reading Drift	—	—	—	—	$V_{IN} = 0V$
		—	0.2	—	—	$0^\circ C \leq T_A \leq +70^\circ C$
		—	1	—	—	$-40^\circ C \leq T_A \leq +85^\circ C$
TC <sub>FS</sub>	Scale Factor Temperature Coefficient	—	—	—	—	$V_{IN} = 399mV$
		—	1	5	ppm/ $^\circ C$	$0^\circ C \leq T_A \leq +70^\circ C$
		—	5	—	ppm/ $^\circ C$	$-40^\circ C \leq T_A \leq +85^\circ C$ Ext Ref = 0ppm/ $^\circ C$
I <sub>S</sub>	Supply Current	—	1	1.5	mA	$V_{IN} = 0V$
	Peak-to-Peak Backplane Drive Voltage	4.25	4.7	5.3	V	$V_S = 9V$ $V_{DISP} = DGND$
	Buzzer Frequency	—	5	—	kHz	$F_{OSC} = 40kHz$
	Counter Time-Base Period	—	1	—	Second	$F_{OSC} = 40kHz$
	Low Battery Flag Voltage	6.7	7	7.3	V	$V_{DD}$ to $V_{SS}$
V <sub>IL</sub>	Input Low Voltage	—	—	DGND + 1.5	V	
V <sub>IH</sub>	Input High Voltage	$V_{DD} - 1.5$	—	—	V	
V <sub>OL</sub>	Output Low Voltage, UR, OR Outputs	$V_{DD} - 1.5$	—	DGND + 0.4	V	$I_L = 50\mu A$
	Control Pin Pull-down Current	—	5	—	$\mu A$	$V_{IN} = V_{DD}$

- Note 1:** Input voltages may exceed the supply voltages provided that input current is limited to  $\pm 100\mu A$ . Current above this value may result in invalid display readings, but will not destroy the device if limited to  $\pm 1mA$ .
- Note 2:** Dissipation ratings assume device is mounted with all leads soldered to printed circuit board.

## 2.0 PIN DESCRIPTIONS

The descriptions of the pins are listed in Table 2-1.

**TABLE 2-1: PIN FUNCTION TABLE**

Pin Number (40-PDIP)	Pin Number (44-PQFP)	Symbol	Description															
1	40	L-E4	LCD segment driver for L ("logic LOW"), polarity, and "e" segment of most significant digit (MSD).															
2	41	AGD4	LCD segment drive for "a," "g," and "d" segments of MSD.															
3	42	BC4P3	LCD segment drive for "b" and "c" segments of MSD and decimal point 3.															
4	43	HFE3	LCD segment drive for H ("logic HIGH"), and "f" and "e" segments of third LSD.															
5	44	AGD3	LCD segment drive for "a," "g," and "d" segments of third LSD.															
6	1	BC3P2	LCD segment drive for "b" and "c" segments of third LSD and decimal point 2.															
7	2	OFE2	LCD segment drive for "over range," and "f" and "e" segments of second LSD.															
8	3	AGD2	LCD segment drive for "a," "g," and "d" segments of second LSD.															
9	4	BC2P1	LCD segment drive for "b " and "c" segments of second LSD and decimal point 1.															
10	5	PKFE1	LCD segment drive for "hold peak reading," and "f" and "e" segments of LSD.															
11	6	AGD1	LCD segment drive for "a," "g," and "d" segments of LSD.															
12	7	BC1BT	LCD segment drive for "b" and "c" segments of LSD and "low battery."															
13	8	BP3	LCD backplane #3.															
14	9	BP2	LCD backplane #2.															
15	10	BP1	LCD backplane #1.															
—	11	V <sub>DISP</sub>	Sets peak LCD drive signal: V <sub>PEAK</sub> = (V <sub>DD</sub> ) – V <sub>DISP</sub> . V <sub>DISP</sub> may also be used to compensate for temperature variation of LCD crystal threshold voltage.															
16	12	DGND	Internal logic digital ground, the logic "0" level. Nominally 4.7V below V <sub>DD</sub> .															
17	13	ANNUNC	Square-wave output at the backplane frequency, synchronized to BP1. ANNUNC can be used to control display annunciators. Connecting an LCD segment to ANNUNC turns it on; connecting it to its backplane turns it off.															
18	14	LOGIC	Logic mode control input. When connected to V <sub>DD</sub> , the converter is in Logic mode. The LCD displays "OL" and the decimal point inputs control the HIGH and LOW annunciators. When the "low" annunciator is on, the buzzer will also be on. When unconnected or connected to DGND, the TC820 is in the Voltage/Frequency Measurement mode. This pin has a 5μA internal pull-down to DGND.															
19	15	RANGE/ FREQ	Dual purpose input. In Range mode, when connected to V <sub>DD</sub> , the integration time will be 200 counts instead of 2000 counts															
20	16	DP0/LO	Dual purpose input. Decimal point select input for voltage measurements. In logic mode, connecting this pin to V <sub>DD</sub> will turn on the "low" LCD segment. There is an internal 5μA pull-down to DGND in Volts mode only. Decimal point logic: <table><tr><td>DP1</td><td>DPQ</td><td>Decimal Point Selected</td></tr><tr><td>0</td><td>0</td><td>None</td></tr><tr><td>0</td><td>1</td><td>DP1</td></tr><tr><td>1</td><td>0</td><td>DP2</td></tr><tr><td>1</td><td>1</td><td>DP3</td></tr></table>	DP1	DPQ	Decimal Point Selected	0	0	None	0	1	DP1	1	0	DP2	1	1	DP3
DP1	DPQ	Decimal Point Selected																
0	0	None																
0	1	DP1																
1	0	DP2																
1	1	DP3																
21	17	DP1/HI	Dual purpose input. Decimal point select input for voltage measurements. In Logic mode, connecting this pin to V <sub>DD</sub> will turn on the "high" LCD segment. There is an internal 5μA pull-down to DGND in Volts mode only.															
22	18	BUZOUT	Buzzer output. Audio frequency, 5kHz, output which drives a piezoelectric buzzer.															
23	19	BUZIN	Buzzer control input. Connecting BUZIN to V <sub>DD</sub> turns the buzzer on. BUZIN is logically OR'ed (internally) with the "logic level low" input. There is an internal 5μA pull-down to DGND.															
24	20	FREQ/ VOLTS	Voltage or frequency measurement select input. When unconnected, or connected VOLTS to DGND, the A/D converter function is active. When connected to V <sub>DD</sub> , the frequency counter function is active. This pin has an internal 5μA pull-down to DGND.															

**TABLE 2-1: PIN FUNCTION TABLE (CONTINUED)**

Pin Number (40-PDIP)	Pin Number (44-PQFP)	Symbol	Description
25	21	PKHOLD	Peak hold input. When connected to $V_{DD}$ , the converter will only update the display if a new conversion value is greater than the preceding value. Thus, the peak reading will be stored and held indefinitely. When unconnected, or connected to DGND, the converter will operate normally. This pin has an internal $5\mu A$ pull-down to DGND.
—	22	UR	Under range output. This output will be HIGH when the digital reading is 380 counts or less.
—	23	OR	Over range output. This output will be HIGH when the analog signal input is greater than full scale. The LCD will display "OL" when the input is over ranged.
26	24	$V_{SS}$	Negative supply connection. Connect to negative terminal of 9V battery.
27	25	COM	Analog circuit ground reference point. Nominally 3.3V below $V_{DD}$ .
28	26	$C_{REF+}$	Positive connection for reference capacitor.
29	27	$C_{REF-}$	Negative connection for reference capacitor.
30	28	$V_{REF+}$	High differential reference input connection.
31	29	$V_{REF-}$	Low differential reference input connection.
32	30	$V_{IN-}$	Low analog input signal connection.
33	31	$V_{IN+}$	High analog input signal connection.
34	32	$V_{BUFF}$	Buffer output. Connect to integration resistor.
35	33	$C_{AZ}$	Auto-zero capacitor connection.
36	34	$V_{INT}$	Integrator output. Connect to integration capacitor.
—	35	$\overline{EOC}/$ HOLD	Bi-directional pin. Pulses low (i.e., from $V_{DD}$ to DGND) at the end of each conversion. If connected to $V_{DD}$ , conversions will continue, but the display is not updated.
37	36	OSC1	Crystal oscillator (input) connection.
38	37	OSC2	Crystal oscillator (output) connection.
39	38	OSC3	RC oscillator connection.
40	39	$V_{DD}$	LCD segment drive for "a," "g," and "d" segments of MSD.

## 3.0 DETAILED DESCRIPTION

The TC820 is a 3-3/4 digit measurement system combining an integrating analog-to-digital converter, frequency counter, and logic level tester in a single package. The TC820 supersedes the TC7106 in new designs by improving performance and reducing system cost. The TC820 adds features that are difficult, expensive, or impossible to provide with older A/D converters (see Table 3-1). The high level of integration permits TC820 based instruments to deliver higher performance and more features, while actually reducing parts count. Fabricated in low power CMOS, the TC820 directly drives a 3-3/4 digit (3999 maximum) LCD.

With a maximum range of 3999 counts, the TC820 provides 10 times greater resolution in the 200mV to 400mV range than traditional 3-1/2 digit meters. An auto-zero cycle ensures a zero reading with a 0V input. CMOS processing reduces analog input bias current to only 1pA. Rollover error (the difference in readings for equal magnitude but opposite polarity input signals) is less than  $\pm 1$  count. Differential reference inputs permit ratiometric measurements for ohms or bridge transducer applications.

The TC820's frequency counter option simplifies design of an instrument well-suited to both analog and digital troubleshooting: voltage, current, and resistance measurements, plus precise frequency measurements to 4MHz (higher frequencies can be measured with an external prescaler), and a simple logic probe. The frequency counter will automatically adjust its range to match the input frequency, over a four-decade range.

Two logic level measurement inputs permit a TC820 based meter to function as a logic probe. When combined with external level shifters, the TC820 will display logic levels on the LCD and also turn on a piezoelectric buzzer when the measured logic level is low.

Other TC820 features simplify instrument design and reduce parts count. On-chip decimal point drivers are included, as is a low battery detection annunciator. A piezoelectric buzzer can be controlled with an external switch or by the logic probe inputs. Two oscillator options are provided: a crystal can be used if high accuracy frequency measurements are desired, or a simple RC option can be used for low-end instruments.

A "peak reading hold" input allows the TC820 to retain the highest A/D or frequency reading. This feature is useful in measuring motor starting current, maximum temperature, and similar applications.

A family of instruments can be created with the TC820. No additional design effort is required to create instruments with 3-3/4 digit resolution.

The TC820 operates from a single 9V battery, with typical power of 10mW. Packages include a 40-pin plastic DIP, 44-pin plastic flat package (PQFP), and 44-pin PLCC.

TABLE 3-1: COMPETITIVE EVALUATION

Features Comparison	TC820	7106
3-3/4 Digit Resolution	Yes	No
Auto-Ranging Frequency Counter	Yes	No
Logic Probe	Yes	No
Decimal Point Drive	Yes	No
Peak Reading Hold (Frequency or Voltage)	Yes	No
Display Hold	Yes	No
Simple 10:1 Range Change	Yes	No
Buzzer Drive	Yes	No
Low Battery Detection with Annunciator	Yes	No
Over Range Detection with Annunciator	Yes	No
Low Drift Reference	Yes	No
Under Range/Over Range Logic Output	Yes	No
Input Overload Display	"OL"	"1"
LCD Annunciator Driver	Yes	No
LCD Drive Type	Triplexed	Direct
LCD Pin Connections	15	24
LCD Elements	36	23

## 3.1 General Theory of Operation

### 3.1.1 DUAL SLOPE CONVERSION PRINCIPLES

The TC820 analog-to-digital converter operates on the principle of dual slope integration. An understanding of the dual slope conversion technique will aid the user in following the detailed TC820 theory of operation following this section. A conventional dual slope converter measurement cycle has two distinct phases:

1. Input Signal Integration
2. Reference Voltage Integration (De-integration)

Referring to Figure 3-1, the unknown input signal to be converted is integrated from zero for a fixed time period ( $t_{INT}$ ), measured by counting clock pulses. A constant reference voltage of the opposite polarity is then integrated until the integrator output voltage returns to zero. The reference integration (de-integration) time ( $t_{DEINT}$ ) is then directly proportional to the unknown input voltage ( $V_{IN}$ ).



In a simple dual slope converter, a complete conversion requires the integrator output to "ramp-up" from zero and "ramp-down" back to zero. A simple mathematical equation relates the input signal, reference voltage, and integration time.

## EQUATION 3-1:

$$\frac{1}{R_{INT}C_{INT}} \int_0^{t_{INT}} V_{IN}(t)dt = \frac{V_{REF}t_{DEINT}}{R_{INT}C_{INT}}$$

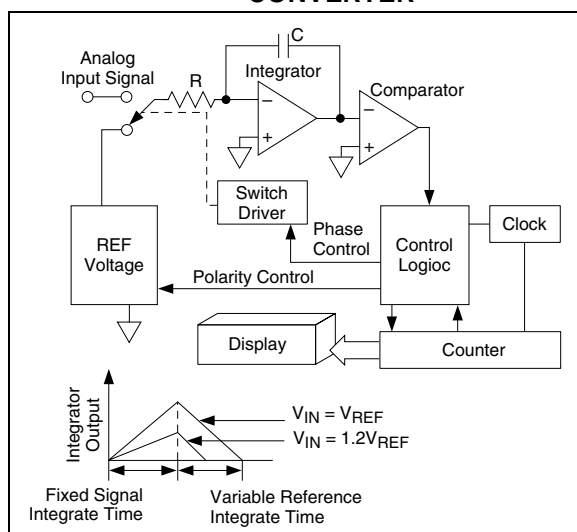
Where:  $V_{REF}$  = Reference Voltage  
 $t_{INT}$  = Integration Time  
 $t_{DEINT}$  = De-integration Time

For a constant  $V_{IN}$ :

## EQUATION 3-2:

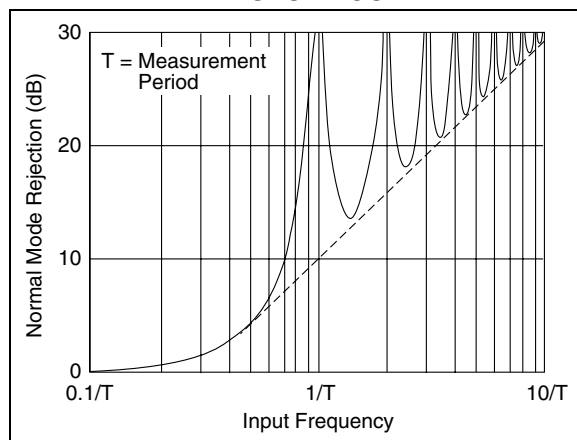
$$V_{IN} = V_{REF} \frac{t_{DEINT}}{t_{INT}}$$

**FIGURE 3-1: BASIC DUAL SLOPE CONVERTER**



Accuracy in a dual slope converter is unrelated to the integrating resistor and capacitor values as long as they are stable during a measurement cycle. An inherent benefit of the dual slope technique is noise immunity. Noise spikes are integrated or averaged to zero during the integration periods, making integrating ADCs immune to the large conversion errors that plague successive approximation converters in high noise environments. Interfering signals, with frequency components at multiples of the averaging (integrating) period, will be attenuated (Figure 3-2). Integrating ADCs commonly operate with the signal integration period set to a multiple of the 50/60Hz power line period.

**FIGURE 3-2: NORMAL MODE REJECTION OF DUAL SLOPE CONVERTER**



## 3.2 Analog Section

In addition to the basic integrate and de-integrate dual slope phases discussed above, the TC820 design incorporates a "zero integrator output" phase and an "auto-zero" phase. These additional phases ensure that the integrator starts at 0V (even after a severe over range conversion), and that all offset voltage errors (buffer amplifier, integrator and comparator) are removed from the conversion. A true digital zero reading is assured without any external adjustments.

A complete conversion consists of four distinct phases:

1. Zero Integrator Output
2. Auto-Zero
3. Signal Integrate
4. Reference De-integrate

### 3.2.1 ZERO INTEGRATOR OUTPUT PHASE

This phase guarantees that the integrator output is at 0V before the system zero phase is entered, ensuring that the true system offset voltages will be compensated for even after an over range conversion. The duration of this phase is 500 counts plus the unused de-integrate counts.

### 3.2.2 AUTO-ZERO PHASE

During the auto-zero phase, the differential input signal is disconnected from the measurement circuit by opening internal analog switches, and the internal nodes are shorted to Analog Common ( $0V_{REF}$ ) to establish a zero input condition. Additional analog switches close a feedback loop around the integrator and comparator to permit comparator offset voltage error compensation. A voltage established on  $C_{AZ}$  then compensates for internal device offset voltages during the measurement cycle. The auto-zero phase residual is typically 10 $\mu$ V to 15 $\mu$ V. The auto-zero duration is 1500 counts.

## 3.2.3 SIGNAL INTEGRATION PHASE

Upon completion of the auto-zero phase, the auto-zero loop is opened and the internal differential inputs connect to  $V_{IN+}$  and  $V_{IN-}$ . The differential input signal is then integrated for a fixed time period, which is 2000 counts (4000 clock periods). The externally set clock frequency is divided by two before clocking the internal counters.

The integration time period is:

**EQUATION 3-3:**

$$t_{INT} = \frac{4000}{F_{OSC}}$$

The differential input voltage must be within the device's Common mode range when the converter and measured system share the same power supply common (ground). If the converter and measured system do not share the same power supply common, as in battery powered applications,  $V_{IN-}$  should be tied to analog common.

Polarity is determined at the end of signal integration phase. The sign bit is a "true polarity" indication, in that signals less than 1LSB are correctly determined. This allows precision null detection that is limited only by device noise and auto-zero residual offsets.

## 3.2.4 REFERENCE INTEGRATE (DE-INTEGRATE) PHASE

The reference capacitor, which was charged during the auto-zero phase, is connected to the input of the integrating amplifier. The internal sign logic ensures the polarity of the reference voltage is always connected in the phase opposite to that of the input voltage. This causes the integrator to ramp back to zero at a constant rate, determined by the reference potential.

The amount of time required ( $t_{DEINT}$ ) for the integrating amplifier to reach zero is directly proportional to the amplitude of the voltage that was put on the integrating capacitor ( $V_{INT}$ ) during the integration phase.

**EQUATION 3-4:**

$$t_{DEINT} = \frac{R_{INT}C_{INT}V_{INT}}{V_{REF}}$$

The digital reading displayed by the TC820 is:

$$\text{Digital Count} = 2000 \frac{V_{IN+} - V_{IN-}}{V_{REF}}$$

The oscillator frequency is divided by 2 prior to clocking the internal decade counters. The four-phase measurement cycle takes a total of 8000 (4000) counts or 16,000 clock pulses. The 8000 count phase is independent of input signal magnitude or polarity.

Each phase of the measurement cycle has the following length:

**TABLE 3-2: MEASUREMENT CYCLE PHASE LENGTH**

Conversion Phase	Counts
1) Auto-Zero	1500
2) Signal Integrate ( <b>Notes 1, 2</b> )	2000
3) Reference Integrate	1 to 4001
4) Integrator Output Zero	499 to 4499

- Note 1:** This time period is fixed. The integration period for the TC820 is:  
 $t_{INT}(\text{TC820}) = 4000/F_{OSC} = 2000 \text{ counts}$ .  
 Where  $F_{OSC}$  is the clock oscillator frequency.
- 2:** Times shown are the RANGE/FREQ at logic low (normal operation). When RANGE/FREQ is logic high, signal integrate times are 200 counts. See Section 3.2.7, "10:1 Range Change".

## 3.2.5 INPUT OVER RANGE

When the analog input is greater than full scale, the LCD will display "OL" and the "OVER RANGE" LCD annunciator will be on.

## 3.2.6 PEAK READING HOLD

The TC820 provides the capability of holding the highest (or peak) reading. Connecting the PK HOLD input to  $V_{DD}$  enables the peak hold feature. At the end of each conversion, the contents of the TC820 counter are compared to the contents of the display register. If the new reading is higher than the reading being displayed, the higher reading is transferred to the display register. A "higher" reading is defined as the reading with the higher absolute value.

The peak reading is held in the display register, so the reading will not "droop" or slowly decay with time. The held reading will be retained until a higher reading occurs, the PK HOLD input is disconnected from  $V_{DD}$ , or power is removed.

The peak signal to be measured must be present during the TC820 signal integrate period. The TC820 does not perform transient peak detection of the analog input signal. However, in many cases, such as measuring temperature or electric motor starting current, the TC820 "acquisition time" will not be a limitation. If true peak detection is required, a simple circuit will suffice. See the applications section for details.

The peak reading function is also available when the TC820 is in the Frequency Counter mode. The counter auto-ranging feature is disabled when peak reading hold is selected.

## 3.2.7 10:1 RANGE CHANGE

The analog input full scale range can be changed with the RANGE/FREQ input. Normally, RANGE/FREQ is held low by an internal pull-down. Connecting this pin to  $V_{S+}$  will increase the full scale voltage by a factor of 10. No external component changes are required.

The RANGE/FREQ input operates by changing the integrate period. When RANGE/FREQ is connected to  $V_{DD}$ , the signal integration phase of the conversion is reduced by a factor of 10 (i.e., from 2000 counts to 200 counts).

For the TC820, the 10:1 range change will result in  $\pm 4V$  full scale. This full scale range will exceed the Common mode range of the input buffer when operating from a 9V battery. If range changing is required for the TC820, a higher supply voltage can be provided, or the input voltage can be divided by 2 externally.

## 3.3 Frequency Counter

In addition to serving as an analog-to-digital converter, the TC820 internal counter can also function as a frequency counter (Figure 3-3). In the Counter mode, pulses at the RANGE/FREQ input will be counted and displayed.

The frequency counter derives its time-base from the clock oscillator. The counter time-base is:

### EQUATION 3-5:

$$t_{\text{COUNT}} = \frac{F_{\text{OSC}}}{40,000}$$

Thus, the counter will operate with a 1-second time-base when a 40kHz oscillator is used. The frequency counter accuracy is determined by the oscillator accuracy. For accurate frequency measurements, a crystal oscillator is recommended.

The frequency counter will automatically select the proper range. Auto-range operation extends over four decades, from 3.999kHz to 3.999MHz. Decimal points are set automatically in the Frequency mode (Figure 3-4). The logic switching levels of the RANGE/FREQ input are CMOS levels. For best counter operation, an external buffer is recommended. See the applications section for details.

## 3.4 Logic Probe

The TC820 can also function as a simple logic probe (Figure 3-5). This mode is selected when the LOGIC input is high. Two dual purpose pins, which normally control the decimal points, are used as logic inputs. Connecting either input to a logic high level will turn on the corresponding LCD annunciator. When the "low" annunciator is on, the buzzer will be on. As with the frequency counter input, external level shifters/buffers are recommended for the logic probe inputs.

FIGURE 3-3: TC820 COUNTER OPERATION

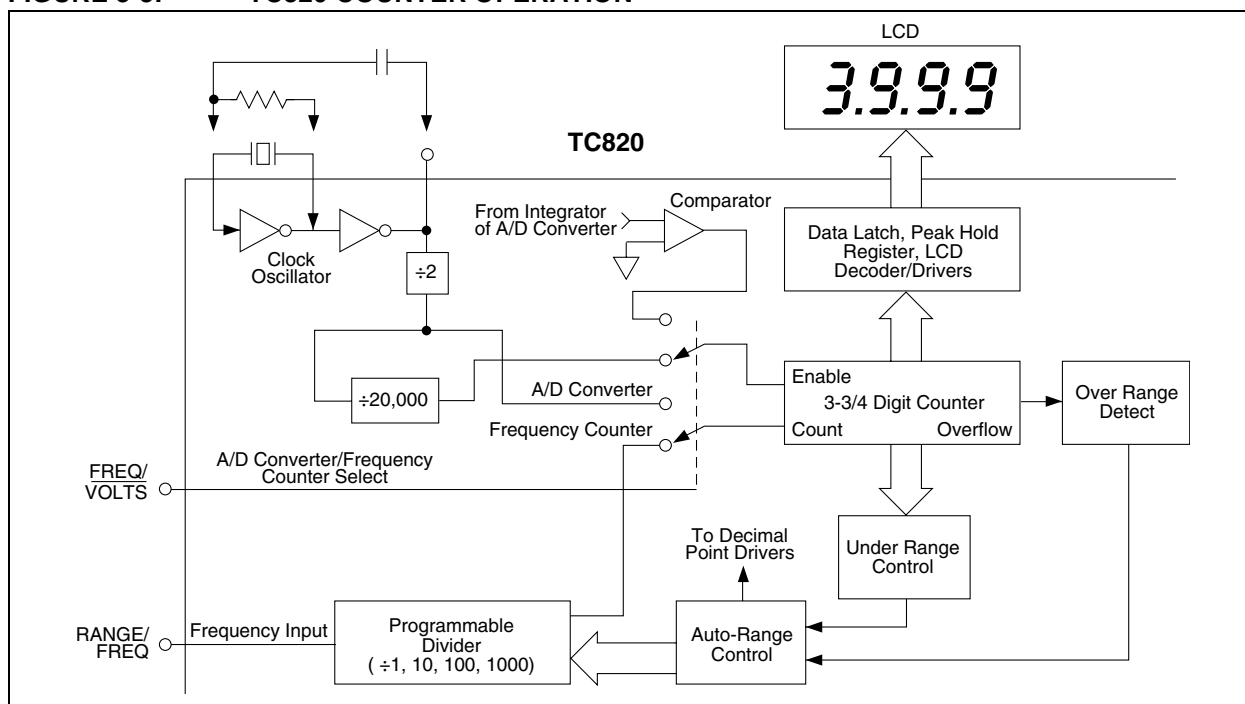


FIGURE 3-4: AUTO-RANGE DECIMAL POINT SELECTION VS. FREQUENCY COUNTER INPUT

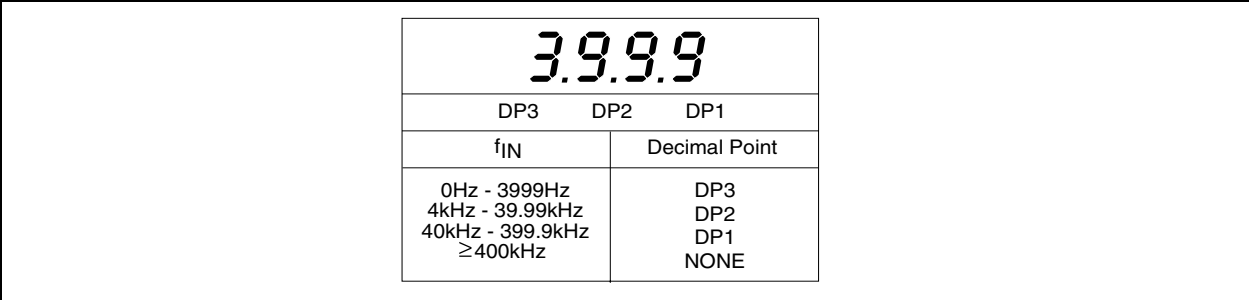
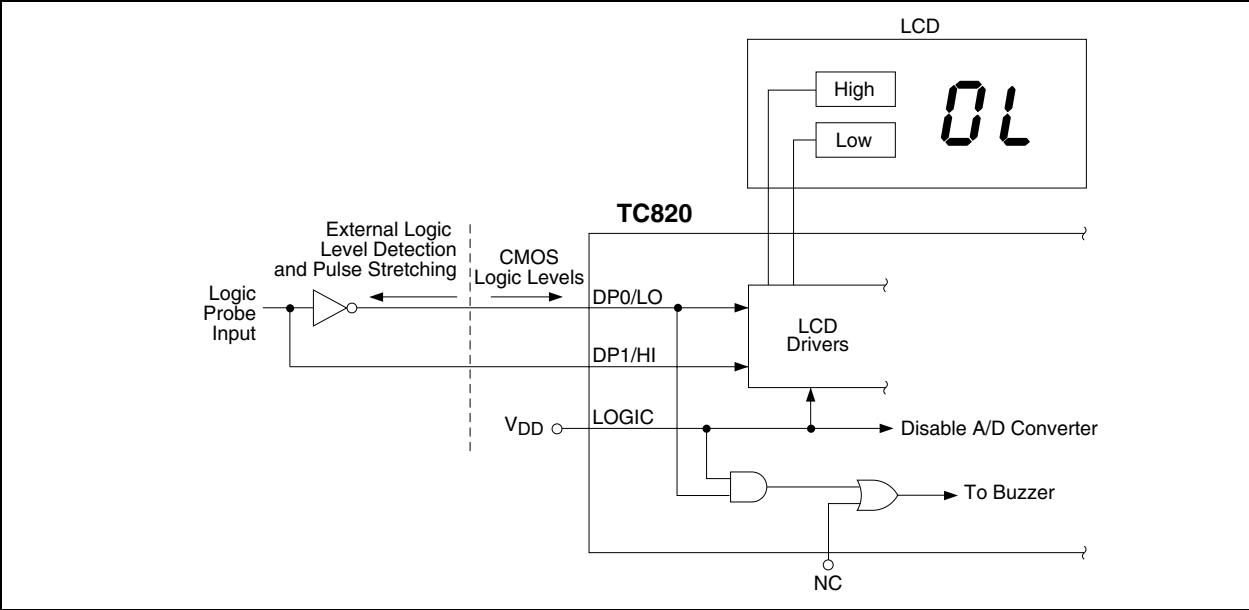
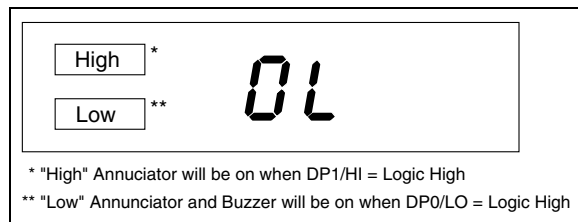


FIGURE 3-5: LOGIC PROBE SIMPLIFIED SCHEMATIC



When the logic probe function is selected while  $FREQ/VOLTS$  is low (A/D mode), the ADC will remain in the Auto-Zero mode. The LCD will read "OL" and all decimal points will be off (Figure 3-6).

**FIGURE 3-6:**



If the logic probe is active while  $FREQ/VOLTS$  is high (Counter mode), the frequency counter will continue to operate. The display will read "OL" but the decimal points will be visible. If the logic probe input is also connected to the RANGE/FREQ input, bringing the LOGIC input low will immediately display the frequency at the logic probe input.

## 3.5 Analog Pin Functional Description

### 3.5.1 DIFFERENTIAL SIGNAL INPUTS ( $V_{IN+}$ ), ( $V_{IN-}$ )

The TC820 is designed with true differential inputs, and accepts input signals within the Input Stage Common mode voltage ( $V_{CM}$ ) range. The typical range is  $V_{DD} - 1V$  to  $V_{SS} + 1.5V$ . Common mode voltages are removed from the system when the TC820 operates from a battery or floating power source (isolated from measured system) and  $V_{SS}$  is connected to analog common (see Figure 3-7).

In systems where Common mode voltages exist, the 86dB Common mode rejection ratio minimizes error. Common mode voltages do, however, affect the integrator output level. A worst case condition exists if a large, positive  $V_{CM}$  exists in conjunction with a full scale, negative differential signal. The negative signal drives the integrator output positive along with  $V_{CM}$  (Figure 3-8). For such applications, the integrator output swing can be reduced below the recommended 2V full scale swing. The integrator output will swing within 0.3V of  $V_{DD}$ , or  $V_{DD}$  without increased linearity error.

### 3.5.2 REFERENCE ( $V_{DD}$ , $V_{SS}$ )

The TC820 reference, like the analog signal input, has true differential inputs. In addition, the reference voltage can be generated anywhere within the power supply voltage of the converter. The differential reference inputs permit ratiometric measurements and simplify interfacing with sensors, such as load cells and temperature sensors.

To prevent rollover type errors from being induced by large Common mode voltages,  $C_{REF}$  should be large compared to stray node capacitance. A 0.1 $\mu$ F capacitor is typical.

The TC820 offers a significantly improved analog common temperature coefficient, providing a very stable voltage suitable for use as a voltage reference. The temperature coefficient of analog common is typically 35ppm/ $^{\circ}$ C.

### 3.5.3 ANALOG COMMON

The analog common pin is set at a voltage potential approximately 3.3V below  $V_{DD}$ . This potential is between 3.15V and 3.45V below  $V_{DD}$ . Analog common is tied internally to an N-channel FET capable of sinking 3mA. This FET will hold the common line at 3.3V below  $V_{DD}$  should an external load attempt to pull the common line toward  $V_{DD}$ . Analog common source current is limited to 12 $\mu$ A, and is, therefore, easily pulled to a more negative voltage (i.e., below  $V_{DD} - 3.3V$ ).

The TC820 connects the internal  $V_{IN+}$  and  $V_{IN-}$  inputs to analog common during the auto-zero cycle. During the reference integrate phase,  $V_{IN-}$  is connected to analog common. If  $V_{IN-}$  is not externally connected to analog common, a Common mode voltage exists.

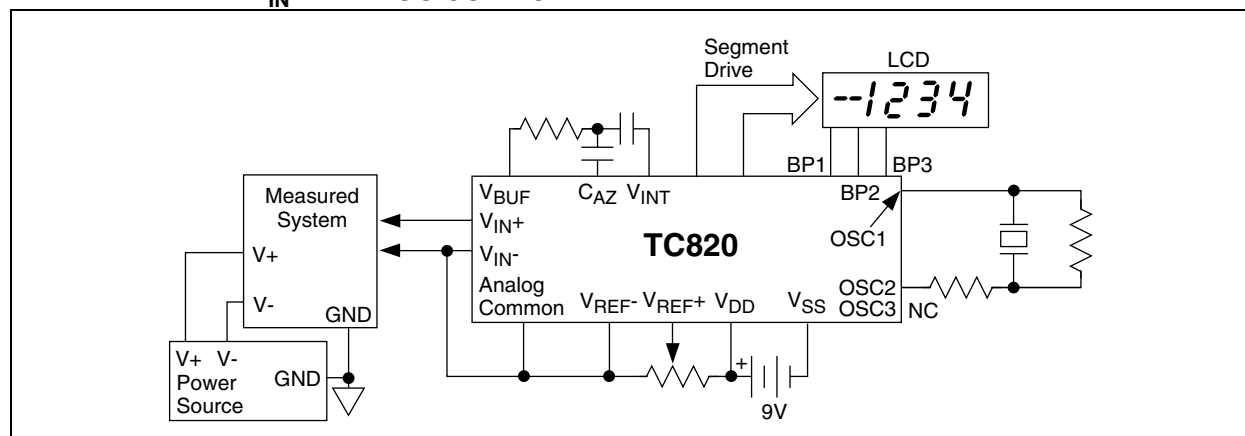
This is rejected by the converter's 86dB Common mode rejection ratio. In battery powered applications, analog common and  $V_{IN-}$  are usually connected, removing Common mode voltage concerns. In systems where  $V_{IN-}$  is connected to the power supply ground or to a given voltage, analog common should be connected to  $V_{IN-}$ .

The analog common pin serves to set the analog section reference or common point. The TC820 is specifically designed to operate from a battery, or in any "measurement" system where input signals are not referenced (float), with respect to the TC820 power source. The analog common potential of  $V_{DD} - 3.3V$  gives a 7V end of battery life voltage. The analog common potential has a voltage coefficient of 0.001%.

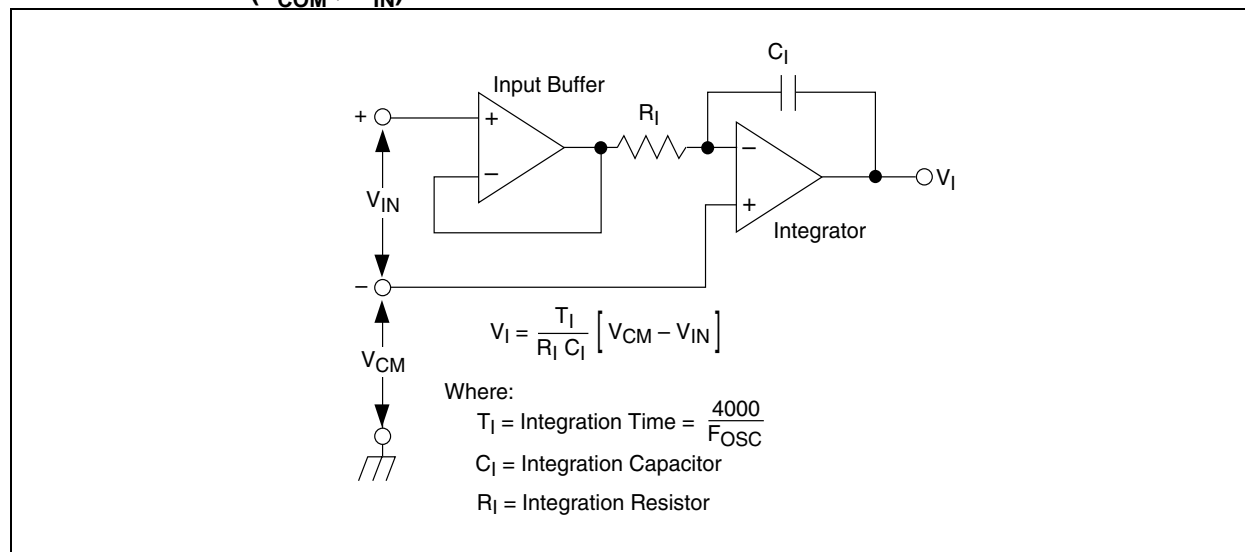
With a sufficiently high total supply voltage ( $V_{DD} - V_{SS} > 7V$ ), analog common is a very stable potential with excellent temperature stability (typically 35ppm/ $^{\circ}$ C). This potential can be used to generate the TC820 reference voltage. An external voltage reference will be unnecessary in most cases, because of the 35ppm/ $^{\circ}$ C temperature coefficient. See the applications section for details.

# TC820

**FIGURE 3-7: COMMON MODE VOLTAGE REMOVED IN BATTERY OPERATION WITH  $V_{IN} = \text{ANALOG COMMON}$**



**FIGURE 3-8: COMMON MODE VOLTAGE REDUCES AVAILABLE INTEGRATOR SWING ( $V_{COM} \neq V_{IN}$ )**



## 4.0 FUNCTION CONTROL INPUTS PIN

### 4.1 Functional Description

The TC820 Operating modes are selected with the function control inputs. See the control input truth, Table 4-1. The high logic threshold is  $\geq V_{DD} - 1.5V$  and the low logic level is  $\leq DGND + 1.5V$ .

**TABLE 4-1: TC820 CONTROL INPUT TRUTH TABLE**

Logic Input			TC820 Function
FREQ/ VOLTS	RANGE/ FREQ	LOGIC	
X	X	1	Logic Probe
0	0	0	A/D Converter, $V_{FULL\ SCALE} = 2 \times V_{REF}$
0	1	0	A/D Converter, $V_{FULL\ SCALE} = 20 \times V_{REF}$
1	Frequency Counter Input	0	Frequency Counter

**Note** 1: Logic "0" = DGND  
2: Logic "1" =  $V_{DD}$

#### 4.1.1 FREQ/VOLTS

This input determines whether the TC820 is in the Analog-to-Digital Conversion mode, or in the Frequency Counter mode. When FREQ/VOLTS is connected to  $V_{DD}$ , the TC820 will measure frequency at the RANGE/FREQ input. When unconnected, or connected to DGND, the TC820 will operate as an analog-to-digital converter. This input has an internal 5 $\mu$ A pull-down to DGND.

#### 4.1.2 LOGIC

The LOGIC input is used to activate the logic probe function. When connected to  $V_{DD}$ , the TC820 will enter the Logic Probe mode. The LCD will show "OL" and all decimal points will be off. The decimal point inputs directly control "high" and "low" display annunciators. When LOGIC is unconnected, or connected to DGND, the TC820 will perform analog-to-digital or frequency measurements, as selected by the FREQ/VOLTS input. The LOGIC input has an internal 5 $\mu$ A pull-down to DGND.

#### 4.1.3 RANGE/FREQ

The function of this dual purpose pin is determined by the FREQ/VOLTS input. When FREQ/VOLTS is connected to  $V_{DD}$ , RANGE/FREQ is the input for the frequency counter function. Pulses at this input are counted with a time-base equal to  $F_{OSC}/40,000$ . Since this input has CMOS input levels ( $V_{DD} - 1.5V$  and  $DGND + 1.5V$ ), an external buffer is recommended.

When the TC820 analog-to-digital converter function is selected, connecting RANGE/FREQ to  $V_{DD}$  will divide the integration time by 10. Therefore, the RANGE/FREQ input can be used to perform a 10:1 range change without changing external components.

#### 4.1.4 DP0/LO, DP1/HI

The function of these dual purpose pins is determined by the LOGIC input. When the TC820 is in the Analog-to-Digital Converter mode, these inputs control the LCD decimal points. See the decimal point truth, Table 4-2. These inputs have internal 5 $\mu$ A pull-downs to DGND when the Voltage/Frequency Measurement mode is active.

**TABLE 4-2: TC820 DECIMAL POINT TRUTH TABLE**

DP1	DP0	LCD
0	0	3999
0	1	399.9
1	0	39.99
1	1	3.999

Connecting the LOGIC input to  $V_{DD}$  places the TC820 in the Logic Probe mode. In this mode, the DP0/LO and DP1/HI inputs control the LCD "low" and "high" annunciators directly. When DP1/HI is connected to  $V_{DD}$ , the "high" annunciator will turn on. When DP0/LO is connected to  $V_{DD}$ , the "low" annunciator and the buzzer will turn on. The internal pull-downs on these pins are disabled when the logic probe function is selected.

These inputs have CMOS logic switching thresholds. For optimum performance as a logic probe, external level shifters are recommended. See the applications section for details.

#### 4.1.5 BUZIN

This input controls the TC820 on-chip buzzer driver. Connecting BUZIN to  $V_{DD}$  will turn the buzzer on. There is an external pull-down to DGND. BUZIN can be used with external circuitry to provide additional functions, such as a fast, audible continuity indication.

## 4.2 Additional Features

The TC820 is available in 40-pin and 44-pin packages. Several additional features are available in the 44-pin package.

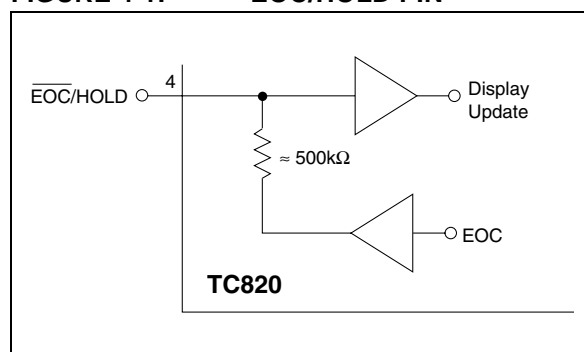
## 4.2.1 $\overline{\text{EOC}}/\text{HOLD}$

$\overline{\text{EOC}}/\text{HOLD}$  is a dual purpose, bi-directional pin. As an output, this pin goes low for 10 clock cycles at the end of each conversion. This pulse latches the conversion data into the display driver section of the TC820.

$\overline{\text{EOC}}/\text{HOLD}$  can be used to hold (or "FREEZE") the display. Connecting this pin to  $V_{DD}$  inhibits the display update process. Conversions will continue, but the display will not change.  $\overline{\text{EOC}}/\text{HOLD}$  will hold the display reading for either analog-to-digital, or frequency measurements.

The input/output structure of the  $\overline{\text{EOC}}/\text{HOLD}$  pin is shown in Figure 4-1. The output drive current is only a few microAmps, so  $\overline{\text{EOC}}/\text{HOLD}$  can easily be overdriven by an open collector logic gate, as well as a FET, bipolar transistor, or mechanical switch. When used as an output,  $\overline{\text{EOC}}/\text{HOLD}$  will have a slow rise and fall time due to the limited output current drive. A CMOS Schmitt trigger buffer is recommended.

**FIGURE 4-1:  $\overline{\text{EOC}}/\text{HOLD}$  PIN**

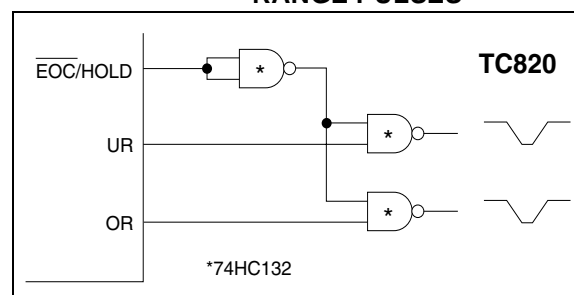


## 4.2.2 OVER RANGE (OR), UNDER RANGE (UR)

The OR output will be high when the analog input signal is greater than full scale (3999 counts). The UR output will be high when the display reading is 380 counts or less.

The OR and UR outputs can be used to provide an auto-ranging meter function. By logically ANDing these outputs with the inverted  $\overline{\text{EOC}}/\text{HOLD}$  output, a single pulse will be generated each time an under ranged or over ranged conversion occurs (Figure 4-2).

**FIGURE 4-2: GENERATING UNDER RANGE AND OVER RANGE PULSES**



## 4.2.3 $V_{\text{DISP}}$

The  $V_{\text{DISP}}$  input sets the peak-to-peak LCD drive voltage. In the 40-pin package,  $V_{\text{DISP}}$  is connected internally to DGND, providing a typical LCD drive voltage of 5V<sub>P-P</sub>. The 44-pin package includes a separate  $V_{\text{DISP}}$  input for applications requiring a variable or temperature compensated LCD drive voltage. See the applications information for suggested circuits.

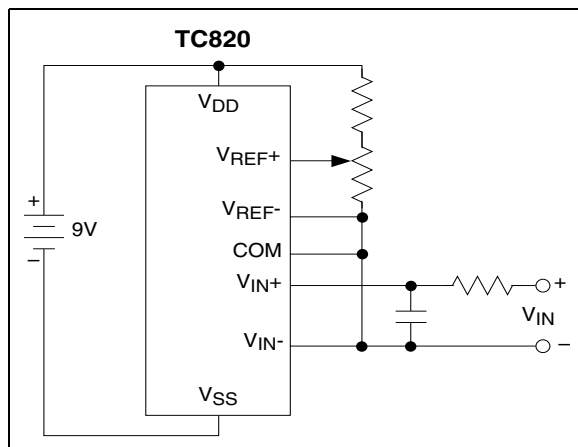


## 5.0 TYPICAL APPLICATIONS

### 5.1 Power Supplies

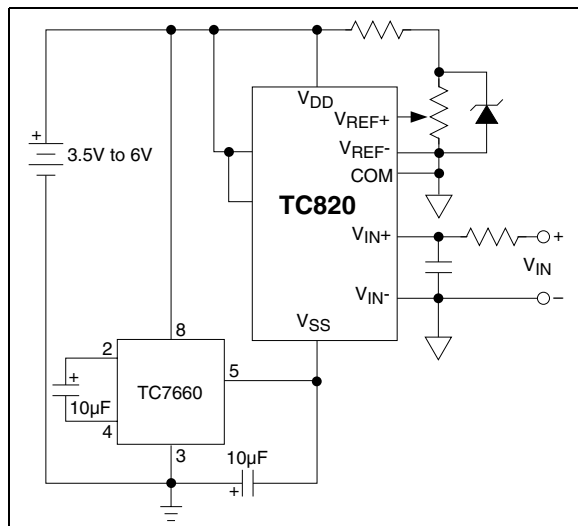
The TC820 is designed to operate from a single power supply such as a 9V battery (Figure 5-1). The converter will operate over a range of 7V to 15V. For battery operation, analog common (COM) provides a Common mode bias voltage (see analog common discussion in the theory of operation section). However, measurements cannot be referenced to battery ground. To do so will exceed the Negative Common mode voltage limit.

**FIGURE 5-1: POWERING THE TC820 FROM A SINGLE 9V BATTERY**



A battery with voltage between 3.5V and 7V can be used to power the TC820, when used with a voltage doubler, as shown in Figure 5-2. The voltage doubler uses the TC7660 and two external capacitors. With this configuration, measurements can be referenced either to analog common or to battery ground.

**FIGURE 5-2: POWERING THE TC820 FROM A LOW VOLTAGE BATTERY**



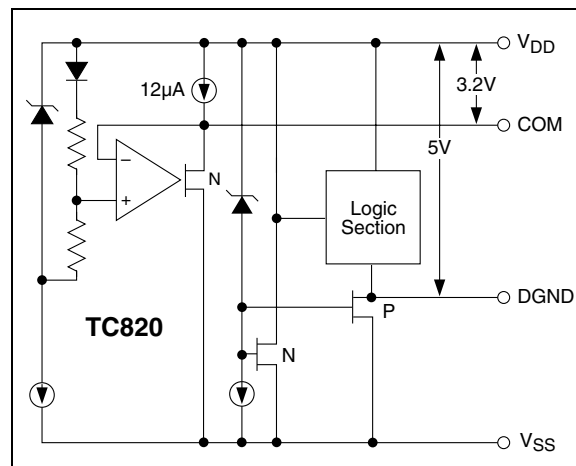
### 5.2 Digital Ground (DGND)

Digital ground is generated from an internal zener diode (Figure 5-3). The voltage between  $V_{DD}$  and DGND is the internal supply voltage for the digital section of the TC820. DGND will sink a minimum of 3mA.

DGND establishes the low logic level reference for the TC820 mode select inputs, and for the frequency and logic probe inputs. The DGND pin can be used as the negative supply for external logic gates, such as the logic probe buffers. To ensure correct counter operation at high frequency, connect a  $1\mu\text{F}$  capacitor from DGND to  $V_{DD}$ .

DGND also provides the drive voltage for the LCD. The TC820 40-pin package internally connects the LCD  $V_{DISP}$  pin to DGND, and provides an LCD drive voltage of about  $5V_{P-P}$ . In the 44-pin package, connecting the  $V_{DISP}$  pin to DGND will provide a 5V LCD drive voltage.

**FIGURE 5-3: DGND AND COM OUTPUTS**



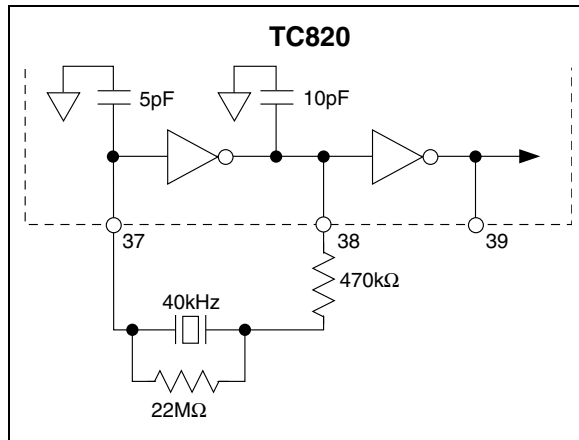
### 5.3 Digital Input Logic Levels

Logic levels for the TC820 digital inputs are referenced to  $V_{DD}$  and DGND. The high level threshold is  $V_{DD} - 1.5V$ , and the low logic level is  $DGND + 1.5V$ . In most cases, digital inputs will be connected directly to  $V_{DD}$  with a mechanical switch. CMOS gates can also be used to control the logic inputs, as shown in the logic probe inputs section.

### 5.4 Clock Oscillator

The TC820 oscillator can be controlled with either a crystal, or with an inexpensive resistor capacitor combination. The crystal circuit, shown in Figure 5-4, is recommended when high accuracy is required in the Frequency Counter mode. The 40kHz crystal is a standard frequency for ultrasonic alarms, and will provide a 1-second time-base for the counter or 2.5 analog-to-digital conversions per second. Consult the crystal manufacturer for detailed applications information.

**FIGURE 5-4: SUGGESTED CRYSTAL OSCILLATOR CIRCUIT**

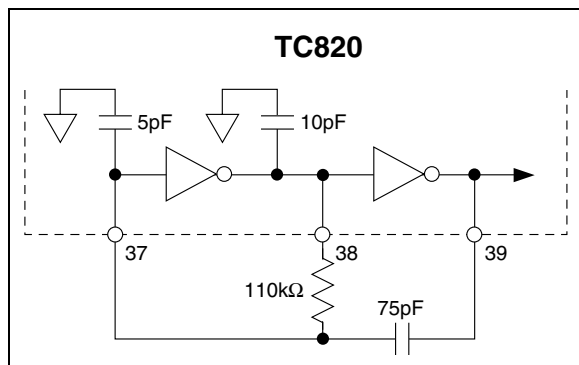


Where low cost is important, the RC circuit of Figure 5-5 can be used. The frequency of this circuit will be approximately:

**EQUATION 5-1:**

$$T_{OSC} = \frac{0.3}{RC}$$

**FIGURE 5-5: RC OSCILLATOR CIRCUIT**

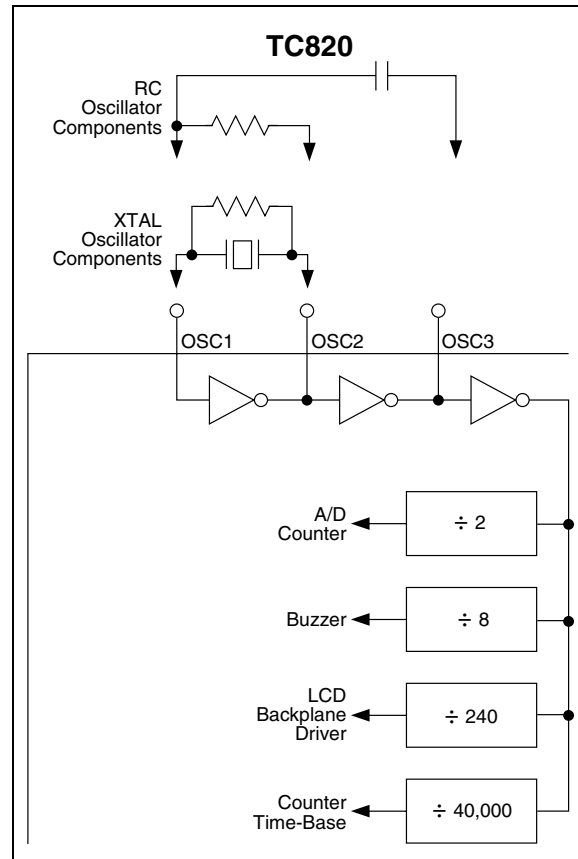


Typical values are  $R = 10k\Omega$  and  $C = 68pF$ . The resistor value should be  $\geq 100k\Omega$ . For accurate frequency measurement, an RC oscillator frequency of 40kHz is required.

## 5.5 System Timing

All system timing is derived from the clock oscillator. The clock oscillator is divided by 2 prior to clocking the A/D counters. The clock is also divided by 8 to drive the buzzer, by 240 to generate the LCD backplane frequency, and by 40,000 for the frequency counter time-base. A simplified diagram of the system clock is shown in Figure 5-6.

**FIGURE 5-6: SYSTEM CLOCK GENERATION**



## 5.6 Component Value Selection

### 5.6.1 AUTO-ZERO CAPACITOR - $C_{AZ}$

The value of the auto-zero capacitor ( $C_{AZ}$ ) has some influence on system noise. A  $0.47\mu F$  capacitor is recommended; a low dielectric absorption capacitor (Mylar) is required.

### 5.6.2 REFERENCE VOLTAGE CAPACITOR - $C_{REF}$

The reference voltage capacitor used to ramp the integrator output voltage back to zero during the reference integrate cycle is stored on  $C_{REF}$ . A  $0.1\mu F$  capacitor is typical. A good quality, low leakage capacitor (such as Mylar) should be used.

## 5.6.3 INTEGRATING CAPACITOR - $C_{INT}$

$C_{INT}$  should be selected to maximize integrator output voltage swing without causing output saturation. Analog common will normally supply the differential voltage reference. For this case, a  $\pm 2V$  integrator output swing is optimum when the analog input is near full scale. For 2.5 readings/second ( $F_{OSC} = 40kHz$ ) and  $V_{FS} = 400mV$ , a  $0.22\mu F$  value is suggested. If a different oscillator frequency is used,  $C_{INT}$  must be changed in inverse proportion to maintain the nominal  $\pm 2V$  integrator swing.

An exact expression for  $C_{INT}$  is:

### EQUATION 5-2:

$$C_{INT} = \frac{4000 V_{FS}}{V_{INT} R_{INT} F_{OSC}}$$

Where:  $F_{OSC}$  = Clock Frequency  
 $V_{FS}$  = Full Scale Input Voltage  
 $R_{INT}$  = Integrating Resistor  
 $V_{INT}$  = Desired Full Scale Integrator Output Swing

$C_{INT}$  must have low dielectric absorption to minimize rollover error. A polypropylene capacitor is recommended.

## 5.6.4 INTEGRATING RESISTOR - $R_{INT}$

The input buffer amplifier and integrator are designed with class A output stages. The integrator and buffer can supply  $40\mu A$  drive currents with negligible linearity errors.  $R_{INT}$  is chosen to remain in the output stage linear drive region, but not so large that printed circuit board leakage currents induce errors. For a  $400mV$  full scale,  $R_{INT}$  should be about  $100k\Omega$ .

## 5.7 Reference Voltage Selection

A full scale reading (4000 counts for TC820) requires the input signal be twice the reference voltage. See Reference Voltage Selection, Table 5-1 below.

**TABLE 5-1: REFERENCE VOLTAGE SELECTION**

Full Scale Input Voltage ( $V_{FS}$ ) (Note 1)	$V_{REF}$	Resolution
200mV	(Note 2)	—
400mV	200mV	$10\mu V$
1V	500mV	$250\mu V$
2V (Notes 3, 4)	1V	$500\mu V$

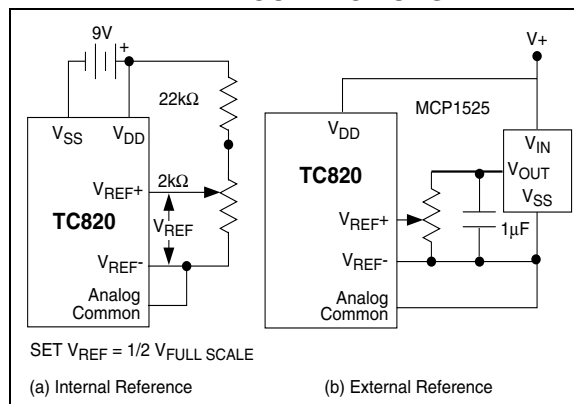
- Note 1:** TC820 in A/D Converter mode, RANGE/FREQ = logic low.  
**Note 2:** Not recommended.  
**Note 3:**  $V_{FS} > 2V$  may exceed the Input Common mode range. See Section 3.2.7, "10:1 Range Change".  
**Note 4:** Full scale voltage values are not limited to the values shown. For example, TC820  $V_{FS}$  can be any value from  $400mV$  to  $2V$ .

In some applications, a scale factor other than unity may exist between a transducer output voltage and the required digital reading. Assume, for example, that a pressure transducer output is  $800mV$  for  $4000 lb/in^2$ . Rather than dividing the input voltage by two, the reference voltage should be set to  $400mV$ . This permits the transducer input to be used directly.

The internal voltage reference potential available at analog common will normally be used to supply the converter's reference voltage. This potential is stable whenever the supply potential is greater than approximately  $7V$ . The low battery detection circuit and analog common operate from the same internal reference. This ensures that the low battery annunciator will turn on at the time the internal reference begins to lose regulation.

The TC820 can also operate with an external reference. Figure 5-7 shows internal and external reference applications.

**FIGURE 5-7: REFERENCE VOLTAGE CONNECTIONS**



## 5.8 Ratiometric Resistance Measurements

The TC820 true differential input and differential reference make ratiometric readings possible. In ratiometric operation, an unknown resistance is measured with respect to a known standard resistance. No accurately defined reference voltage is needed.

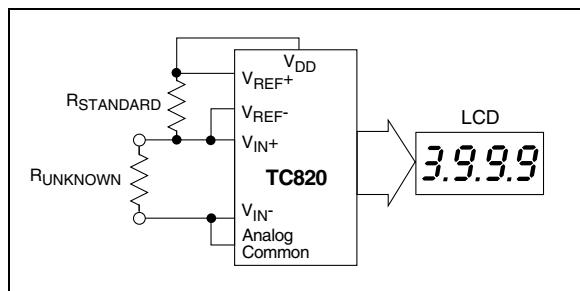
The unknown resistance is put in series with a known standard and a current is passed through the pair (Figure 5-8). The voltage developed across the unknown is applied to the input and voltages across the known resistor applied to the reference input. If the unknown equals the standard, the input voltage will equal the reference voltage and the display will read 2000. The displayed reading can be determined from the following expression:

### EQUATION 5-3:

$$\text{Displayed Reading} = \frac{R_{UNKNOWN}}{R_{STANDARD}}$$

The display will over range for values of  $R_{UNKNOWN} \geq 2 \times R_{STANDARD}$ .

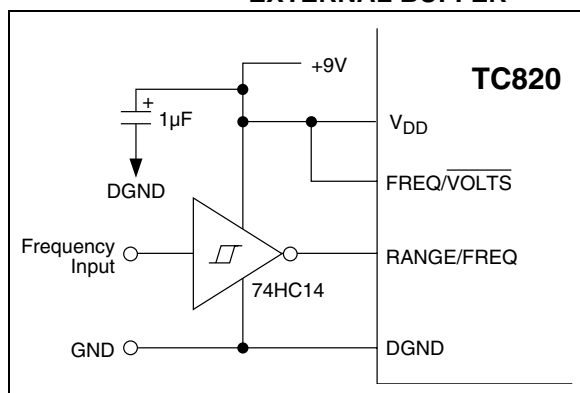
**FIGURE 5-8: LOW PARTS COUNT RATIO-METRIC RESISTANCE MEASUREMENT**



## 5.9 Buffering the FREQ Input

When the FREQ/VOLTS input is high and the LOGIC input is low, the TC820 will count pulses at the RANGE/FREQ input. The time-base will be  $F_{OSC}/40,000$ , or 1 second with a 40kHz clock. The signal to be measured should swing from  $V_{DD}$  to DGND. The RANGE/FREQ input has CMOS input levels without hysteresis. For best results, especially with low frequency sine-wave inputs, an external buffer with hysteresis should be added. A typical circuit is shown in Figure 5-9.

**FIGURE 5-9: FREQUENCY COUNTER EXTERNAL BUFFER**

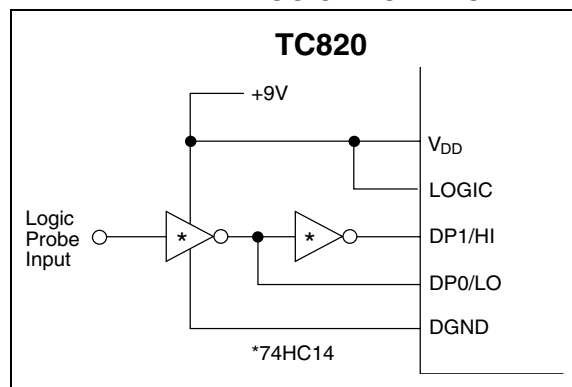


## 5.10 Logic Probe Inputs

The DP0/LO and DP1/HI inputs provide the logic probe inputs when the LOGIC input is high. Driving either DP0/LO or DP1/HI to a logic high will turn on the appropriate LCD annunciator. When DP0/LO is high, the buzzer will be on.

To provide a "single input" logic probe function, external buffers should be used. A simple circuit is shown in Figure 5-10. This circuit will turn the appropriate annunciator on for high and low level inputs.

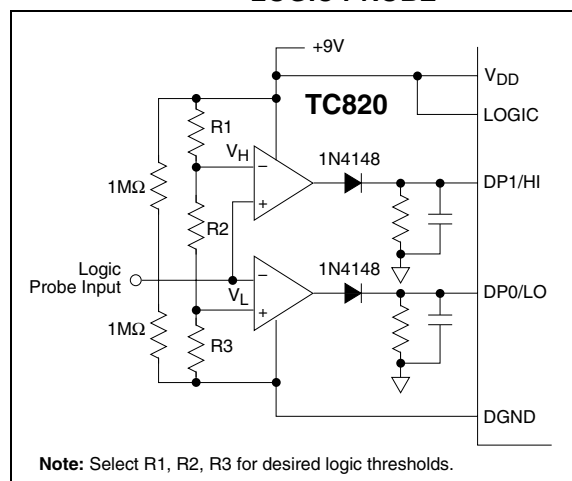
**FIGURE 5-10: SIMPLE EXTERNAL LOGIC PROBE BUFFER**



If carefully controlled logic thresholds are required, a window comparator can be used. Figure 5-11 shows a typical circuit. This circuit will turn on the high or low annunciators when the logic thresholds are exceeded, but the resistors connected from DP0/LO and DP1/HI to DGND will turn both annunciators off when the logic probe is unconnected.

The TC820 logic inputs are not latched internally, so pulses of short duration will usually be difficult or impossible to see. To display short pulses properly, the input pulse should be "stretched." The circuit of Figure 5-11 shows capacitors added across the input pull-down resistors to stretch the input pulse and permit viewing short duration input pulses.

**FIGURE 5-11: WINDOW COMPARATOR LOGIC PROBE**

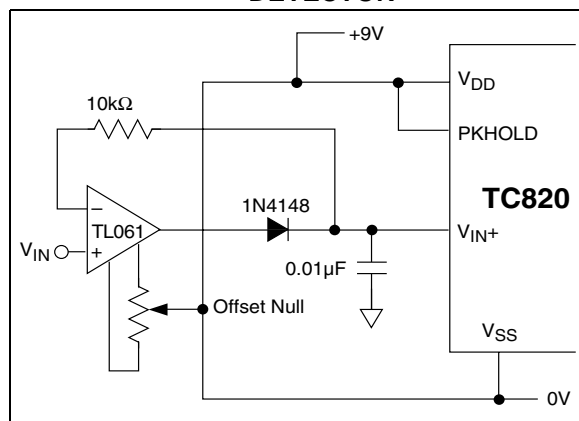


## 5.11 External Peak Detection

The TC820 will hold the highest A/D conversion or frequency reading indefinitely when the PKHOLD input is connected to  $V_{DD}$ . However, the analog peak input must be present during the A/D converter's signal integrate period. For slowly changing signals, such as temperature, the peak reading will be properly converted and held.

If rapidly changing analog signals must be held, an external peak detector should be added. An inexpensive circuit can be made from an op amp and a few discrete components, as shown in Figure 5-12. The droop rate of the external peak detector should be adjusted so that the held voltage will not decay below the desired accuracy level during the converter's 400msec conversion time.

**FIGURE 5-12: EXTERNAL PEAK DETECTOR**

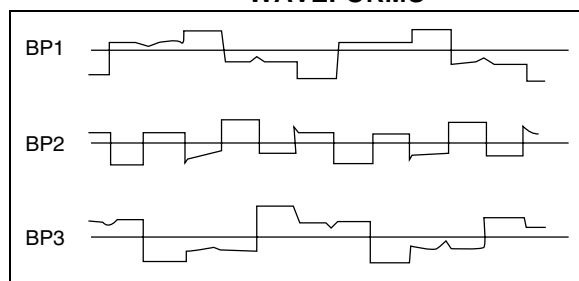


## 5.12 Liquid Crystal Display (LCD)

The TC820 drives a triplex (multiplexed 3:1) LCD with three backplanes. The LCD can include decimal points, polarity sign, and annunciators for over range, peak hold, high and low logic levels, and low battery. Table 5-2 shows the assignment of the display segments to the backplanes and segment drive lines. The backplane drive frequency is obtained by dividing the oscillator frequency by 240.

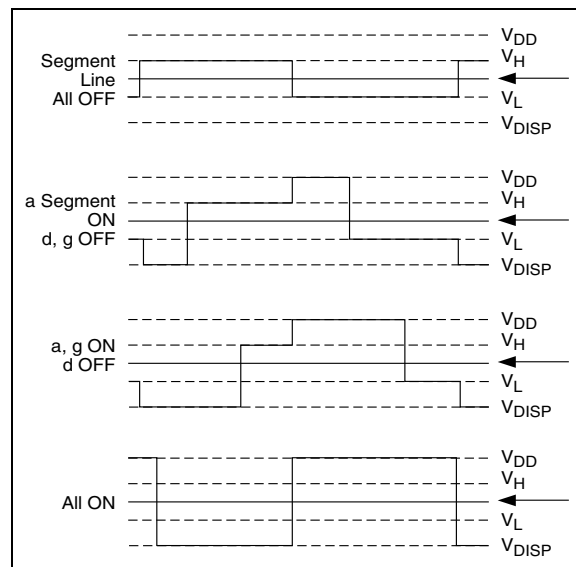
Backplane waveforms are shown in Figure 5-13. These appear on outputs BP1, BP2, and BP3. They remain the same, regardless of the segments being driven.

**FIGURE 5-13: BACKPLANE WAVEFORMS**



Other display output lines have waveforms that vary depending on the displays values. Figure 5-13 shows a set of waveforms for the a, g, d outputs of one digit for several combinations of "on" segments.

**FIGURE 5-14: TYPICAL DISPLAY OUTPUT WAVEFORMS**



**TABLE 5-2: LCD BACKPLANE AND SEGMENT ASSIGNMENTS**

40-Pin (PDIP)	44-Pin (PQFP)	LCD Display Pin Number	BP1	BP2	BP3
1	40	3	LOW		
2	41	4	A4	"—"	
3	42	5	B4	G4	E4
4	43	6	HIGH	C4	D4
5	44	7	A3	F3	DP3
6	1	8	B3	G3	E3
7	2	9	OVER	C3	D3
8	3	10	A2	F2	DP2
9	4	11	B2	G2	E2
10	5	12	PEAK	C2	D2
11	6	13	A1	F1	DP1
12	7	14	B1	G1	E1
13	8	2,16*	—	C1	D1
14	9	1	BP1	—	BATT
15	10	15	LOW	BP2	BP3

\*Connect both pins 2 and 16 of LCD to TC820 BP3 of output.

## 5.13 LCD Source

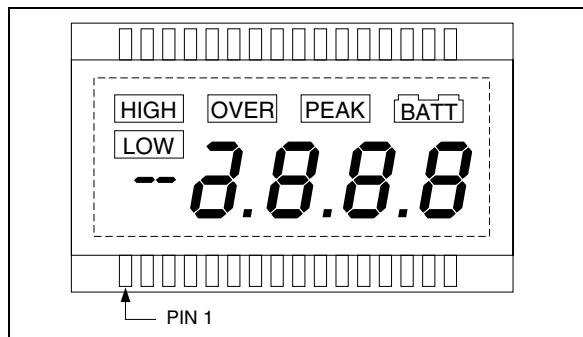
Although most users will design their own custom LCD, a standard display for the TC820 (Figure 5-15), Part No. ST-1355-M1, is available from the following sources:

Crystaloid (USA)  
Crystaloid Electronics  
P.O. Box 628

5282 Hudson Drive  
Hudson, OH 44238  
Phone: 216-655-2429

Crystaloid (Europe)  
Rep. France  
102, rue des Nouvelles  
F92150 Suresnes France  
Phone: 33-1-42-04-29-25  
Fax: 33-1-45-06-46-99

**FIGURE 5-15: TYPICAL TC820 LCD**



## 5.14 Annunciator Output

The annunciator output is a square wave running at the backplane frequency (for example, 167Hz when  $F_{OSC} = 40\text{kHz}$ ). The peak-to-peak amplitude is equal to  $(V_{DD} - V_{DISP})$ . Connecting an annunciator of the LCD to the annunciator output turns it on; connecting it to its backplane turns it off.

## 5.15 LCD Drive Voltage ( $V_{DISP}$ )

The peak-to-peak LCD drive voltage is equal to  $(V_{DD} - V_{DISP})$ . In the 40-pin dual in-line package (DIP),  $V_{DISP}$  is internally connected to DGND, providing a typical LCD drive voltage of  $5V_{P-P}$ .

For applications with a wide temperature range, some LCDs require that the drive levels vary with temperature to maintain good viewing angle and display contrast. In this case, the TC820 44-pin package provides a pin connection for  $V_{DISP}$ . Figure 5-16 shows TC820 circuits that can be adjusted to give a temperature compensation of about  $10\text{mV}/^{\circ}\text{C}$  between  $V_{DD}$  and  $V_{DISP}$ . The diode between GND and  $V_{DISP}$  should have a low turn on voltage because  $V_{DISP}$  cannot exceed  $0.3\text{V}$  below GND.

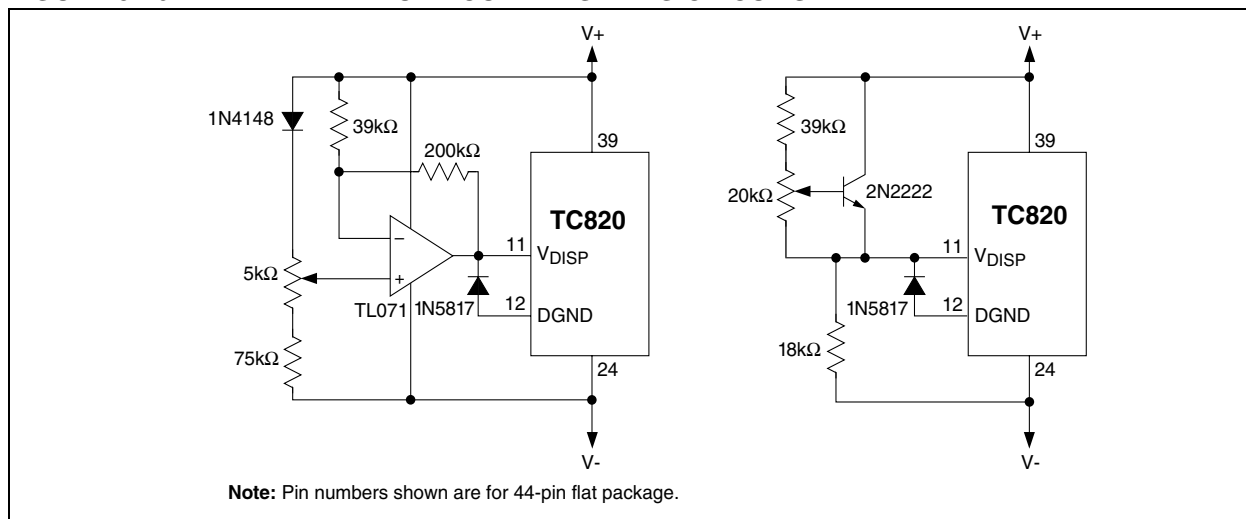
## 5.16 Crystal Source

Two sources of the 40kHz crystal are:

Statek Corp.  
512 N. Main St.  
Orange, CA 92668  
Phone: 714-639-7810  
Fax: 714-997-1256  
Part #: CX-1V-40.0

SPK Electronics  
2F-1, No. 312, Sec. 4, Jen Ai Rd.  
Taipei, Taiwan R.O.C.  
Phone: (02) 754-2677  
Fax: 886-2-708-4124  
Part #: QRT-38-40.0kHz

**FIGURE 5-16: TEMPERATURE COMPENSATING CIRCUITS**



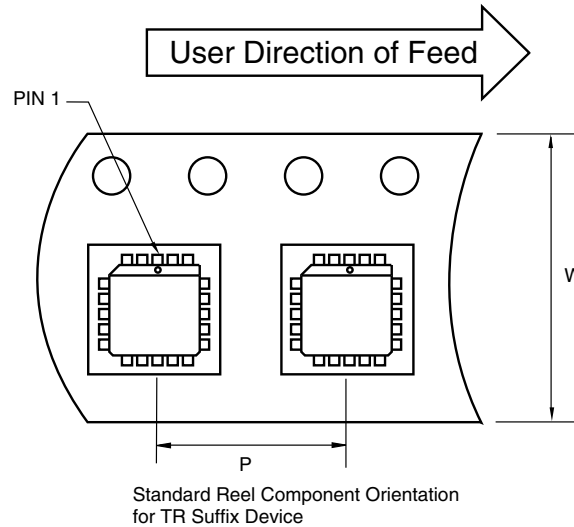
## 6.0 PACKAGING INFORMATION

### 6.1 Package Marking Information

Package marking data not available at this time.

### 6.2 Taping Forms

#### Component Taping Orientation for 44-Pin PLCC Devices

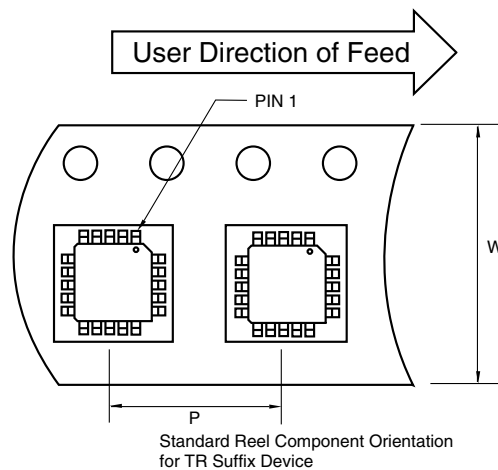


#### Carrier Tape, Number of Components Per Reel and Reel Size

Package	Carrier Width (W)	Pitch (P)	Part Per Full Reel	Reel Size
44-Pin PLCC	32 mm	24 mm	500	13 in

**Note:** Drawing does not represent total number of pins.

#### Component Taping Orientation for 44-Pin PQFP Devices



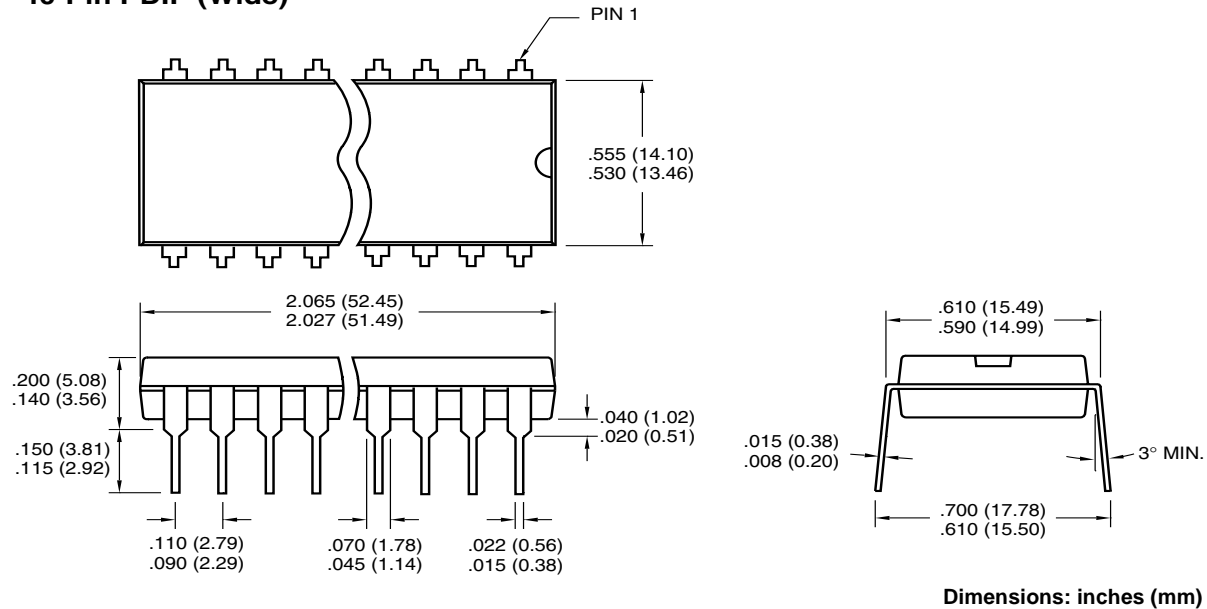
#### Carrier Tape, Number of Components Per Reel and Reel Size

Package	Carrier Width (W)	Pitch (P)	Part Per Full Reel	Reel Size
44-Pin PQFP	24 mm	16 mm	500	13 in

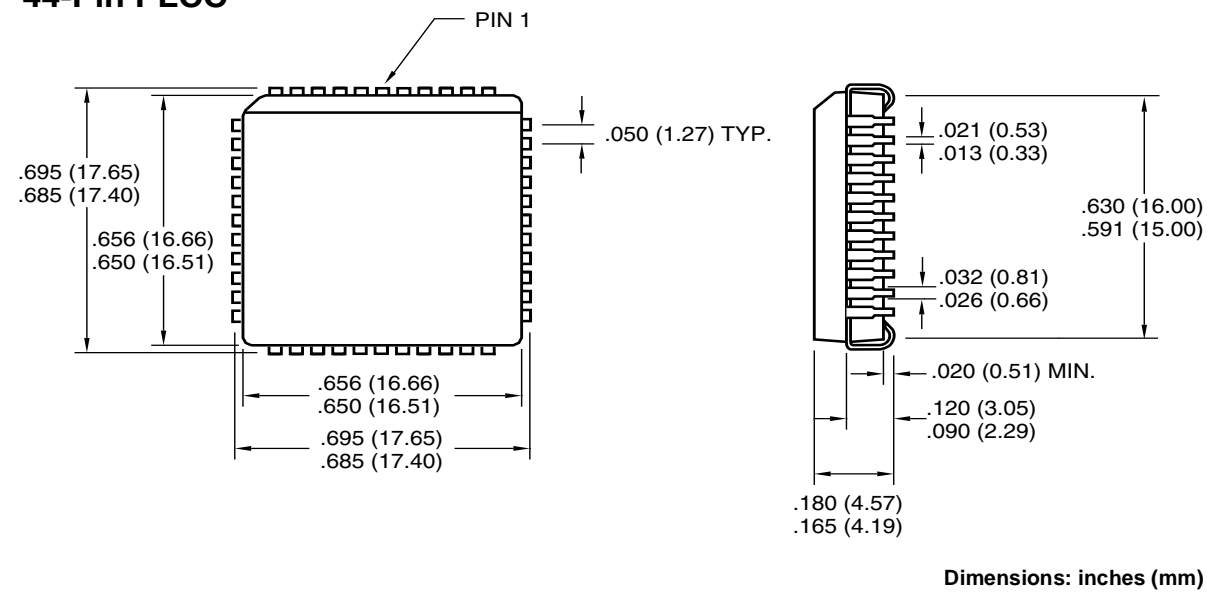
**Note:** Drawing does not represent total number of pins.

## 6.3 Package Dimensions

### 40-Pin PDIP (Wide)



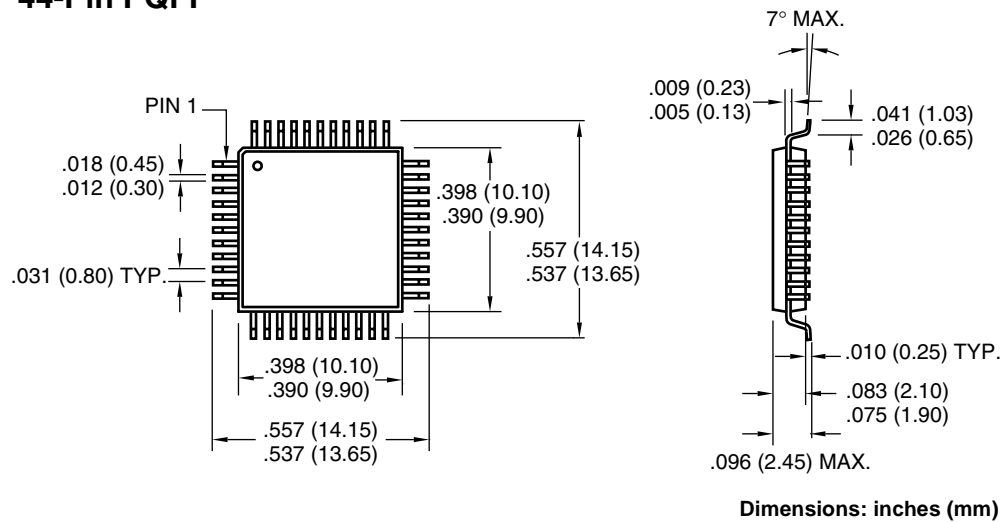
### 44-Pin PLCC





## 6.3 Package Dimensions (Continued)

### 44-Pin PQFP



## SALES AND SUPPORT

### **Data Sheets**

Products supported by a preliminary Data Sheet may have an errata sheet describing minor operational differences and recommended workarounds. To determine if an errata sheet exists for a particular device, please contact one of the following:

1. Your local Microchip sales office
2. The Microchip Corporate Literature Center U.S. FAX: (480) 792-7277
3. The Microchip Worldwide Site ([www.microchip.com](http://www.microchip.com))

Please specify which device, revision of silicon and Data Sheet (include Literature #) you are using.

### **New Customer Notification System**

Register on our web site ([www.microchip.com/cn](http://www.microchip.com/cn)) to receive the most current information on our products.

Information contained in this publication regarding device applications and the like is intended through suggestion only and may be superseded by updates. It is your responsibility to ensure that your application meets with your specifications. No representation or warranty is given and no liability is assumed by Microchip Technology Incorporated with respect to the accuracy or use of such information, or infringement of patents or other intellectual property rights arising from such use or otherwise. Use of Microchip's products as critical components in life support systems is not authorized except with express written approval by Microchip. No licenses are conveyed, implicitly or otherwise, under any intellectual property rights.

## Trademarks


The Microchip name and logo, the Microchip logo, FilterLab, KEELOQ, microID, MPLAB, PIC, PICmicro, PICMASTER, PICSTART, PRO MATE, SEEVAL and The Embedded Control Solutions Company are registered trademarks of Microchip Technology Incorporated in the U.S.A. and other countries.

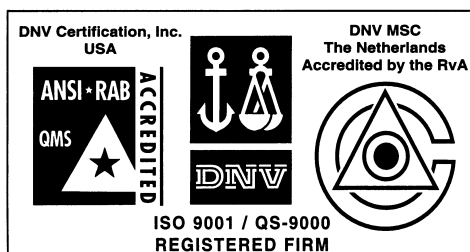
dsPIC, ECONOMONITOR, FanSense, FlexROM, fuzzyLAB, In-Circuit Serial Programming, ICSP, ICEPIC, microPort, Migratable Memory, MPASM, MPLIB, MPLINK, MPSIM, MXDEV, PICC, PICDEM, PICDEM.net, rfPIC, Select Mode and Total Endurance are trademarks of Microchip Technology Incorporated in the U.S.A.

Serialized Quick Turn Programming (SQTP) is a service mark of Microchip Technology Incorporated in the U.S.A.

All other trademarks mentioned herein are property of their respective companies.

© 2002, Microchip Technology Incorporated, Printed in the U.S.A., All Rights Reserved.

 Printed on recycled paper.



*Microchip received QS-9000 quality system certification for its worldwide headquarters, design and wafer fabrication facilities in Chandler and Tempe, Arizona in July 1999 and Mountain View, California in March 2002. The Company's quality system processes and procedures are QS-9000 compliant for its PICmicro® 8-bit MCUs, KEELOQ® code hopping devices, Serial EEPROMs, microperipherals, non-volatile memory and analog products. In addition, Microchip's quality system for the design and manufacture of development systems is ISO 9001 certified.*



## WORLDWIDE SALES AND SERVICE

### AMERICAS

#### Corporate Office

2355 West Chandler Blvd.  
Chandler, AZ 85224-6199  
Tel: 480-792-7200 Fax: 480-792-7277  
Technical Support: 480-792-7627  
Web Address: <http://www.microchip.com>

#### Rocky Mountain

2355 West Chandler Blvd.  
Chandler, AZ 85224-6199  
Tel: 480-792-7966 Fax: 480-792-7456

#### Atlanta

500 Sugar Mill Road, Suite 200B  
Atlanta, GA 30350  
Tel: 770-640-0034 Fax: 770-640-0307

#### Boston

2 Lan Drive, Suite 120  
Westford, MA 01886  
Tel: 978-692-3848 Fax: 978-692-3821

#### Chicago

333 Pierce Road, Suite 180  
Itasca, IL 60143  
Tel: 630-285-0071 Fax: 630-285-0075

#### Dallas

4570 Westgrove Drive, Suite 160  
Addison, TX 75001  
Tel: 972-818-7423 Fax: 972-818-2924

#### Detroit

Tri-Atria Office Building  
32255 Northwestern Highway, Suite 190  
Farmington Hills, MI 48334  
Tel: 248-538-2250 Fax: 248-538-2260

#### Kokomo

2767 S. Albright Road  
Kokomo, Indiana 46902  
Tel: 765-864-8360 Fax: 765-864-8387

#### Los Angeles

18201 Von Karman, Suite 1090  
Irvine, CA 92612  
Tel: 949-263-1888 Fax: 949-263-1338

#### New York

150 Motor Parkway, Suite 202  
Hauppauge, NY 11788  
Tel: 631-273-5305 Fax: 631-273-5335

#### San Jose

Microchip Technology Inc.  
2107 North First Street, Suite 590  
San Jose, CA 95131  
Tel: 408-436-7950 Fax: 408-436-7955

#### Toronto

6285 Northam Drive, Suite 108  
Mississauga, Ontario L4V 1X5, Canada  
Tel: 905-673-0699 Fax: 905-673-6509

### ASIA/PACIFIC

#### Australia

Microchip Technology Australia Pty Ltd  
Suite 22, 41 Rawson Street  
Epping 2121, NSW  
Australia  
Tel: 61-2-9868-6733 Fax: 61-2-9868-6755

#### China - Beijing

Microchip Technology Consulting (Shanghai)  
Co., Ltd., Beijing Liaison Office  
Unit 915  
Bei Hai Wan Tai Bldg.  
No. 6 Chaoyangmen Beidajie  
Beijing, 100027, No. China  
Tel: 86-10-85282100 Fax: 86-10-85282104

#### China - Chengdu

Microchip Technology Consulting (Shanghai)  
Co., Ltd., Chengdu Liaison Office  
Rm. 2401, 24th Floor,  
Ming Xing Financial Tower  
No. 88 TIDU Street  
Chengdu 610016, China  
Tel: 86-28-6766200 Fax: 86-28-6766599

#### China - Fuzhou

Microchip Technology Consulting (Shanghai)  
Co., Ltd., Fuzhou Liaison Office  
Unit 28F, World Trade Plaza  
No. 71 Wusi Road  
Fuzhou 350001, China  
Tel: 86-591-7503506 Fax: 86-591-7503521

#### China - Shanghai

Microchip Technology Consulting (Shanghai)  
Co., Ltd.  
Room 701, Bldg. B  
Far East International Plaza  
No. 317 Xian Xia Road  
Shanghai, 200051  
Tel: 86-21-6275-5700 Fax: 86-21-6275-5060

#### China - Shenzhen

Microchip Technology Consulting (Shanghai)  
Co., Ltd., Shenzhen Liaison Office  
Rm. 1315, 13/F, Shenzhen Kerry Centre,  
Renminnan Lu  
Shenzhen 518001, China  
Tel: 86-755-2350361 Fax: 86-755-2366086

#### Hong Kong

Microchip Technology Hongkong Ltd.  
Unit 901-6, Tower 2, Metroplaza  
223 Hing Fong Road  
Kwai Fong, N.T., Hong Kong  
Tel: 852-2401-1200 Fax: 852-2401-3431

#### India

Microchip Technology Inc.  
India Liaison Office  
Divyasree Chambers  
1 Floor, Wing A (A3/A4)  
No. 11, O'Shaugnessey Road  
Bangalore, 560 025, India  
Tel: 91-80-2290061 Fax: 91-80-2290062

### Japan

Microchip Technology Japan K.K.  
Benex S-1 6F  
3-18-20, Shinyokohama  
Kohoku-Ku, Yokohama-shi  
Kanagawa, 222-0033, Japan  
Tel: 81-45-471-6166 Fax: 81-45-471-6122

### Korea

Microchip Technology Korea  
168-1, Youngbo Bldg. 3 Floor  
Samsung-Dong, Kangnam-Ku  
Seoul, Korea 135-882  
Tel: 82-2-554-7200 Fax: 82-2-558-5934

### Singapore

Microchip Technology Singapore Pte Ltd.  
200 Middle Road  
#07-02 Prime Centre  
Singapore, 188980  
Tel: 65-6334-8870 Fax: 65-6334-8850

### Taiwan

Microchip Technology Taiwan  
11F-3, No. 207  
Tung Hua North Road  
Taipei, 105, Taiwan  
Tel: 886-2-2717-7175 Fax: 886-2-2545-0139

### EUROPE

#### Denmark

Microchip Technology Nordic ApS  
Regus Business Centre  
Lautrup høj 1-3  
Ballerup DK-2750 Denmark  
Tel: 45 4420 9895 Fax: 45 4420 9910

#### France

Microchip Technology SARL  
Parc d'Activite du Moulin de Massy  
43 Rue du Saule Trapu  
Batiment A - 1er Etage  
91300 Massy, France  
Tel: 33-1-69-53-63-20 Fax: 33-1-69-30-90-79

#### Germany

Microchip Technology GmbH  
Gustav-Heinemann Ring 125  
D-81739 Munich, Germany  
Tel: 49-89-627-144 0 Fax: 49-89-627-144-44

#### Italy

Microchip Technology SRL  
Centro Direzionale Colleoni  
Palazzo Taurus 1 V. Le Colleoni 1  
20041 Agrate Brianza  
Milan, Italy  
Tel: 39-039-65791-1 Fax: 39-039-6899883

#### United Kingdom

Arizona Microchip Technology Ltd.  
505 Eskdale Road  
Winnersh Triangle  
Wokingham  
Berkshire, England RG41 5TU  
Tel: 44 118 921 5869 Fax: 44-118 921-5820

03/01/02

